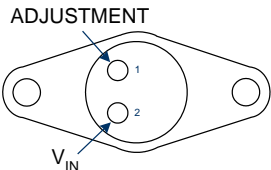
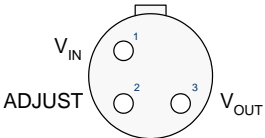
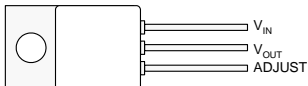
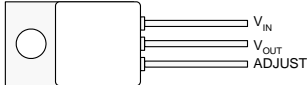




**THE DATASHEET OF
SG117HVIG-DESC**



Connection Diagrams and Ordering Information

Ambient Temperature	Type	Package	Part Number	Packaging Type	Connection Diagram
-55°C to 125°C	K	3-TERMINAL METAL CAN	SG117AK	TO-3	
			SG117AK-883B		
			SG117AK-DESC		
			SG117K		
			SG117K-883B		
			SG117K-DESC		
			SG117K-JAN		
-55°C to 125°C	T	3-TERMINAL METAL CAN	SG117AT	TO-39	
			SG117AT-883B		
			SG117AT-DESC		
			SG117T		
			SG117T-883B		
			SG117T-DESC		
			SG117T-JAN		
-55°C to 125°C	IG	3-Pin HERMETIC Package (Isolated)	SG117AIG	TO-257	
			SG117AIG-883B		
			SG117AIG-DESC		
			SG117IG		
			SG117IG-883B		
			SG117IG-DESC		
-55°C to 125°C	G	3-Pin HERMETIC Package	SG117AG	TO-257	
			SG117AG-883B		
			SG117AG-DESC		
			SG117G		
			SG117G-883B		
			SG117G-DESC		

Absolute Maximum Ratings

Parameter	Value	Units
Power Dissipation	Internally Limited	
Input to Output Voltage Differential	40	V
Storage Temperature Range	-65 to 150	°C
Maximum Operating Junction Temperature	150	°C
Lead Temperature (Soldering, 10 seconds)	300	°C

Notes:
Exceeding these ratings could cause damage to the device. All voltages are with respect to Ground. Currents are positive into, negative out of specified terminal.

Thermal Data

Parameter	Value	Units
K Package: 3 Terminal TO-3 Metal Can		
Thermal Resistance-Junction to Case, θ_{JC}	3	°C/W
Thermal Resistance-Junction to Ambient, θ_{JA}	35	°C/W
T Package: 3-Pin TO-39 Metal Can		
Thermal Resistance-Junction to Case, θ_{JC}	15	°C/W
Thermal Resistance-Junction to Ambient, θ_{JA}	120	°C/W
IG Package: 3-Pin TO-257 Hermetic (Isolated)		
Thermal Resistance-Junction to Case, θ_{JC}	3.5	°C/W
Thermal Resistance-Junction to Ambient, θ_{JA}	42	°C/W
G Package: 3-Pin TO-257 Hermetic		
Thermal Resistance-Junction to Case, θ_{JC}	3.5	°C/W
Thermal Resistance-Junction to Ambient, θ_{JA}	42	°C/W
L Package: 20-Pin Ceramic (LCC) Leadless		
Thermal Resistance-Junction to Case, θ_{JC}	35	°C/W
Thermal Resistance-Junction to Ambient, θ_{JA}	120	°C/W

Notes:

- Junction Temperature Calculation: $T_J = T_A + (P_D \times \theta_{JA})$.
- The above numbers for θ_{JC} are maximums for the limiting thermal resistance of the package in a standard mounting configuration. The θ_{JA} numbers are meant to be guidelines for the thermal performance of the device/pcboard system. All of the above assume no ambient airflow.

Recommended Operating Conditions

Parameter	Value	Units
Input Voltage Range	$V_{OUT} + 3.5V$ to 37	V
Operating Ambient Temperature Range		
SG117A / SG117	-55 to 125	°C

Note: Range over which the device is functional.

Electrical Characteristics

Unless otherwise specified, these characteristics apply over the full operating ambient temperature for the SG117A / SG117 with $-55^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$, $V_{\text{IN}} - V_{\text{OUT}} = 5.0\text{V}$ and for $I_{\text{OUT}} = 500\text{mA}$ (K, G, and IG) and $I_{\text{OUT}} = 100\text{mA}$ (T, and L packages). Although power dissipation is internally limited, these specifications are applicable for power dissipations of 2W for the T, and L packages, and 20W for the K, G, and IG packages. I_{MAX} is 1.5A for the K, G, and IG packages and 500mA for the T, and L packages. Low duty cycle pulse testing techniques are used which maintains junction and case temperatures equal to the ambient temperature.

Parameter	Test Condition	SG117A			SG117			Units	
		Min	Typ	Max	Min	Typ	Max		
Reference Section									
Reference Voltage	$I_{\text{OUT}} = 10\text{mA}$, $T_A = 25^{\circ}\text{C}$	1.238	1.250	1.262				V	
	$3\text{V} \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 40\text{V}$, $P \leq P_{\text{MAX}}$	1.225	1.250	1.270	1.20	1.25	1.30		
	$10\text{mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$								
Output Section									
Line Regulation ¹	$3\text{V} \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 40\text{V}$, $I_L = 10\text{mA}$							%V	
	$T_A = 25^{\circ}\text{C}$		0.005	0.01		0.01	0.02		
	$T_A = T_{\text{MIN}}$ to T_{MAX}		0.01	0.02		0.02	0.05		
Load Regulation ¹	$10\text{mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$							mV	
	$V_{\text{OUT}} \leq 5\text{V}$, $T_A = 25^{\circ}\text{C}$		5	15		5	15		
	$V_{\text{OUT}} > 5\text{V}$, $T_A = 25^{\circ}\text{C}$		0.1	0.3		0.1	0.3		%
	$V_{\text{OUT}} \leq 5\text{V}$		20	50		20	50		mV
	$V_{\text{OUT}} > 5\text{V}$		0.3	1		0.3	1		%
Thermal Regulation ²	$T_A = 25^{\circ}\text{C}$, 20ms pulse		0.002	0.02		0.03	0.07	%/W	
Ripple Rejection	$V_{\text{OUT}} = 10\text{V}$, $f = 120\text{Hz}$							dB	
	$C_{\text{ADJ}} = 1\mu\text{F}$, $T_A = 25^{\circ}\text{C}$		65			65			
	$C_{\text{ADJ}} = 10\mu\text{F}$	66	80		66	80			
Minimum Load Current	$(V_{\text{IN}} - V_{\text{OUT}}) = 40\text{V}$		3.5	5		3.5	5	mA	
Current Limit	$(V_{\text{IN}} - V_{\text{OUT}}) \leq 15\text{V}$							A	
	K, P, G, IG Packages	1.5	2.2		1.5	2.2			
	T, L Packages	0.5	0.8		0.5	0.8			
	$(V_{\text{IN}} - V_{\text{OUT}}) = 40\text{V}$, $T_J = 25^{\circ}\text{C}$							A	
	K, P, G, IG Packages	0.3	0.4		0.3	0.4			
	T, L Packages	0.15	0.2		0.15	0.2			
Temperature Stability ²			1	2		1		%	
Long Term Stability ²	$T_A = 125^{\circ}\text{C}$, 1000 Hours		0.3	1		0.3	1		
RMS Output Noise (% of V_{OUT}) ²	$T_A = 25^{\circ}\text{C}$, $10\text{Hz} \leq f \leq 10\text{kHz}$		0.001			0.001			
Adjust Section									
Adjust Pin Current			50	100		50	100	μA	
Adjust Pin Current Change	$10\text{mA} < I_{\text{OUT}} < I_{\text{MAX}}$, $2.5\text{V} < (V_{\text{IN}} - V_{\text{OUT}}) < 40\text{V}$		0.2	5		0.2	5		
¹ Regulation is measured at constant junction temperature, using pulse testing with low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation. ² These parameters, although guaranteed, are not tested in production.									

Characteristic Curves

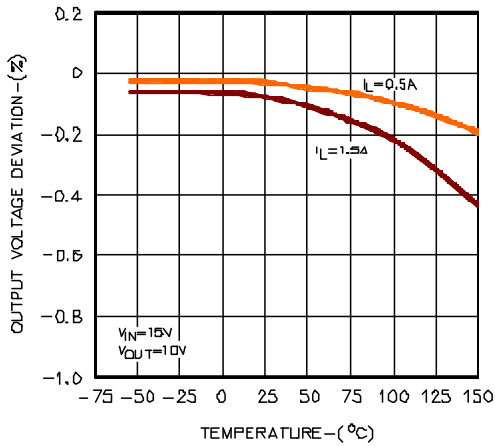


Figure 2 - Output Voltage Deviation vs. Temperature

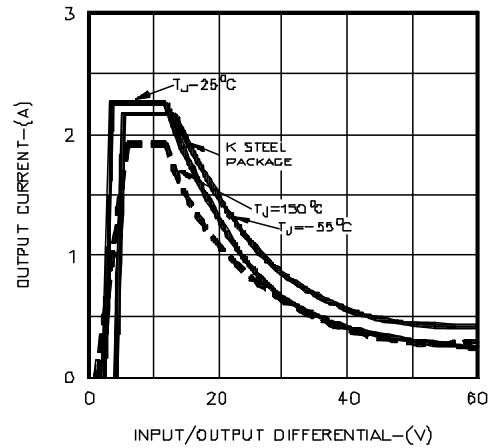


Figure 3 - Output Current vs. Input / Output Differential

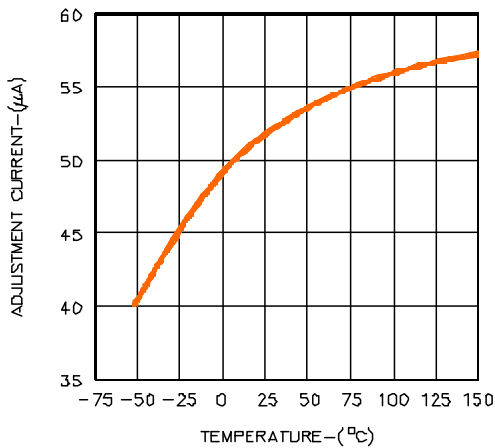


Figure 4 - Adjust Current vs. Temperature

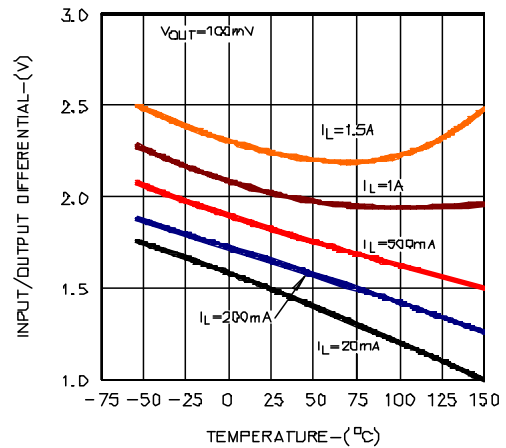


Figure 5 - Input / Output Differential vs. Temperature

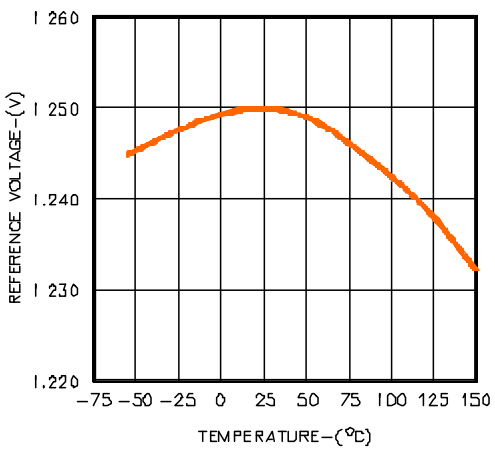


Figure 6 - Reference Voltage vs. Temperature

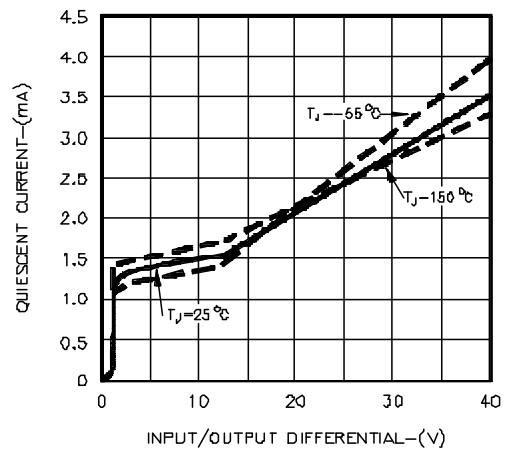


Figure 7 - Quiescent Current vs. Input / Output Differential

Characteristic Curves

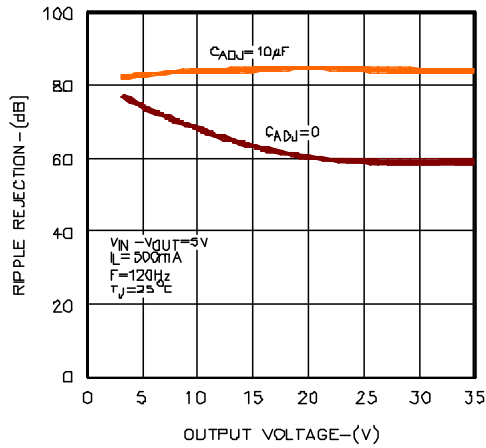


Figure 8 - Ripple Rejection vs. Output Voltage

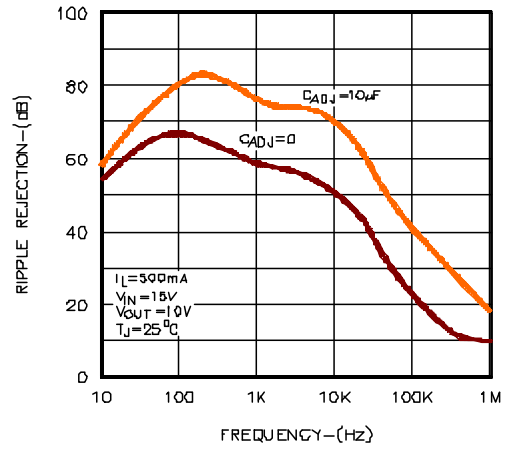


Figure 9 - Ripple Rejection vs. Frequency

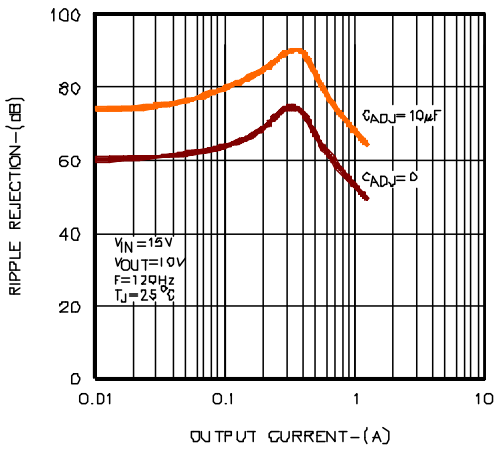


Figure 10 - Ripple Rejection vs. Output Current

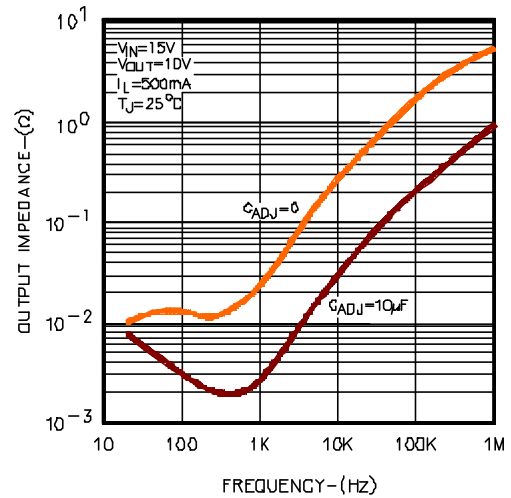


Figure 11 - Output Impedance vs. Frequency

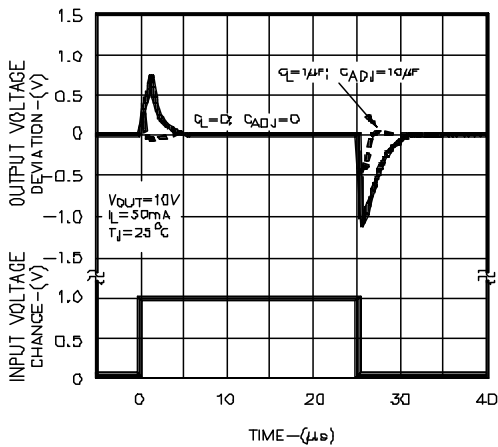


Figure 12 - Line Transient Response

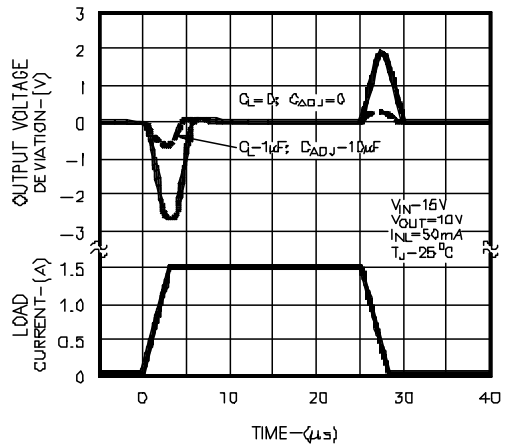


Figure 13 - Load Transient Response

Characteristic Curves

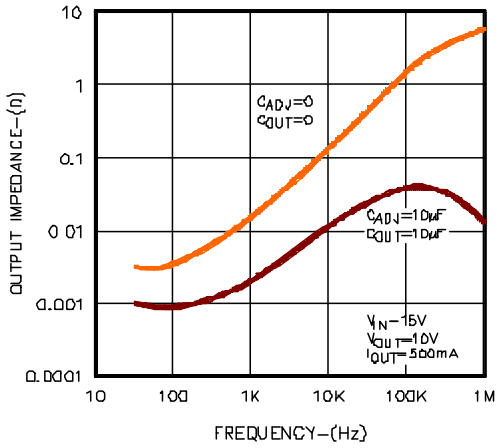


Figure 14 · Output Impedance vs. Frequency

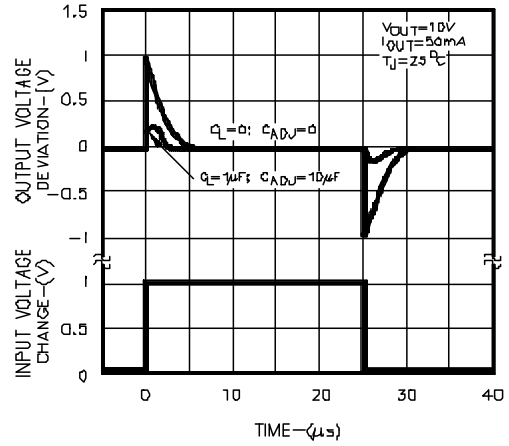


Figure 15 · Line Transient Response

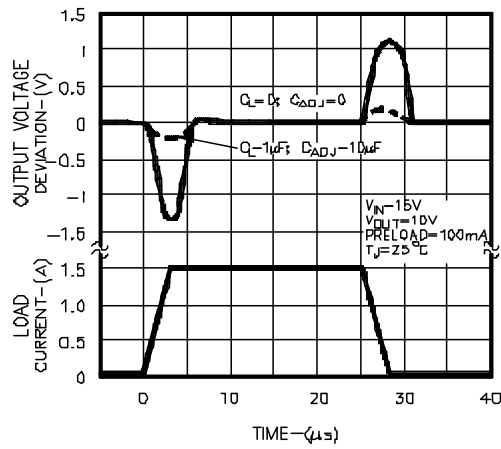


Figure 16 · Load Transient Response

Application Information

General

The SG117A develops a 1.25V reference voltage between the output (OUT) and the adjust (ADJ) terminals (see Basic Regulator Circuit). By placing a resistor, R_1 between these two terminals, a constant current is caused to flow through R_1 and down through R_2 to set the overall output voltage. Normally this current is the specified minimum load current of 5mA or 10mA. It is important to maintain this minimum output load current requirement otherwise the device may fail to regulate, and the output voltage may rise.

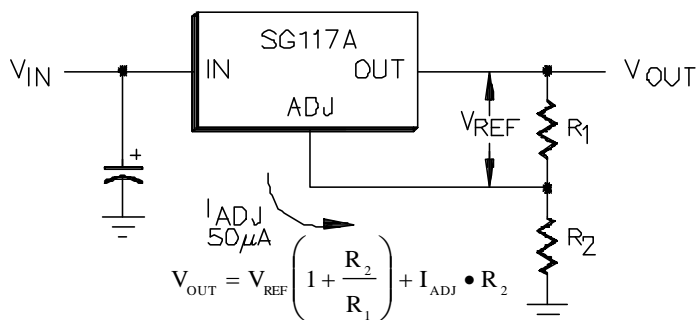


Figure 17 - Basic Regulator Circuit

The I_{ADJ} current does add an error to the output divider ratio, however because I_{ADJ} is very small and constant when compared with the current through R_1 , it represents a small error and can often be ignored.

It is easily seen from the above equation, that even if the resistors were of exact value, the accuracy of the output is limited by the accuracy of V_{REF} . With a guaranteed 1% reference, a 5V power supply design, using $\pm 2\%$ resistors, would have a worst case manufacturing tolerance of $\pm 4\%$. If 1% resistors were used, the tolerance would drop to $\pm 2.5\%$. A plot of the worst case output voltage tolerance as a function of resistor tolerance is shown below.

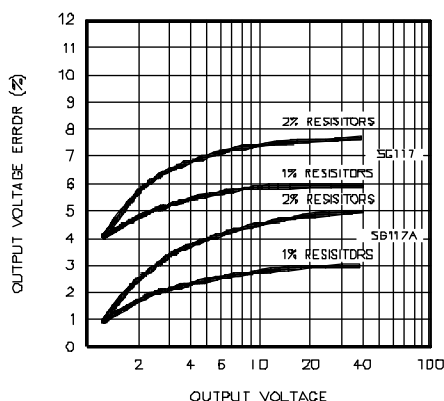
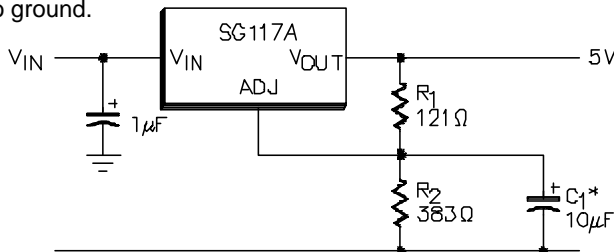


Figure 18 - Voltage Tolerance vs. Resistor Tolerance

Bypass Capacitors

Input bypassing using a 0.1 μF ceramic or 1 μF solid tantalum is recommended, and especially when any input filter capacitors are more than 5 inches from the device. A 0.1 μF bypass capacitor on the ADJ pin is required if the load current varies by more than 1A/ μsec . Improved ripple rejection (80dB) can be accomplished by adding a 10 μF capacitor from the ADJ pin to ground.



* C_1 Improves Ripple Rejection. X_C should be small compared to R_2 .

Figure 19 - Improving Ripple Rejection

While the SG117 is stable with no output capacitor, for improved AC transient response and to prevent the possibility of oscillation due to an unknown reactive load, a $1\mu\text{F}$ capacitor is also recommended at the output. Because of their low impedance at high frequencies, the best type of capacitor to use is solid tantalum; ceramic capacitors may also be used. When bypass capacitors are used, it may be necessary to provide external protection diodes to prevent this external large capacitance from discharging through internal low current paths, which may damage the device. Although the duration of any surge current is short, there may be sufficient energy to damage the regulator. This is particularly true of the large capacitance on the ADJ pin when output voltages are higher than 25V. Such a capacitor could discharge into the ADJ pin when either the input or output is shorted. See figure below.

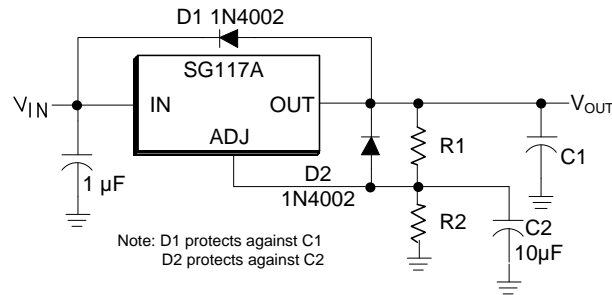


Figure 20 - Use of Protection Diodes

Load Regulation

Because the SG117A is a three-terminal device, it is not possible to provide true remote load sensing. Load regulation will be limited by the resistance of the wire connecting the regulator to the load. From the data sheet specification, regulation is measured at the bottom of the package. Negative side sensing is a true Kelvin connection, with the bottom of the output divider returned to the negative side of the load. Although it may not be immediately obvious, best load regulation is obtained when the top of the divider is connected directly to the case, not to the load. This is illustrated in (Connections for Best Load Regulation). If R1 were connected to the load, the effective resistance between the regulator and the load would be:

$$R_p \cdot \left(\frac{R_2 + R_1}{R_1} \right), R_p = \text{Parasitic Line Resistance}$$

Connected as shown, R_p is not multiplied by the divider ratio. R_p is about 0.004Ω per foot using 16 gauge wire. This translates to 4mV/ft. at 1A load current, so it is important to keep the positive lead between regulator and load as short as possible.

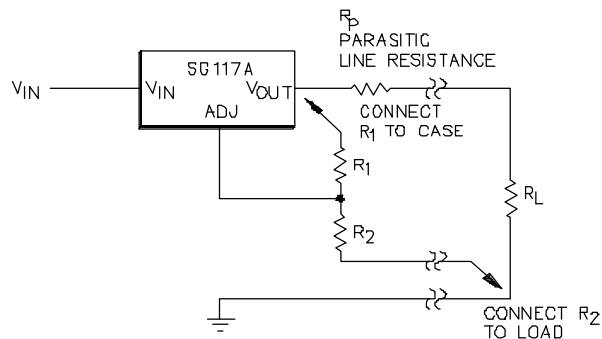


Figure 21 - Connections for Best Load Regulation

Current Limit

As outlined in the Electrical Characteristics the current limit will activate whenever the output current exceeds the specified levels. It is also important to bear in mind that the regulator includes a foldback-current characteristic that limits the current at higher V_{IN} to V_{OUT} differential voltages. This power limiting characteristic will prevent the regulator from providing full output current depending on the V_{IN} to V_{OUT} differential. Also if during a short circuit situation the regulator was presented with a voltage that exceeds the Absolute Maximum Rating of 40V (e.g. $V_{IN} > 40V$, $V_{OUT} = 0V$) the device may fail, or be permanently damaged.

Typical Applications

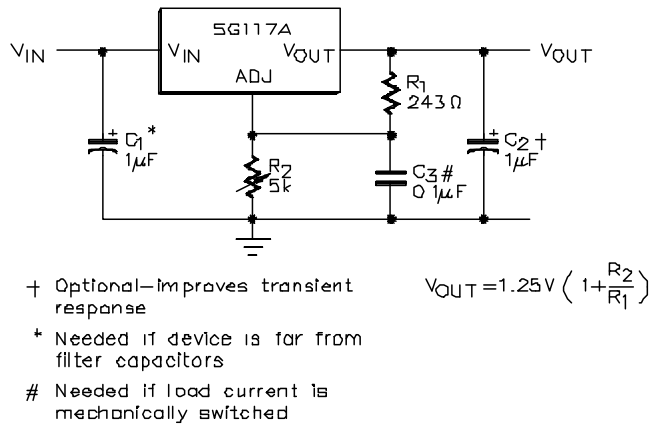


Figure 22 - 1.2V – 25V Adjustable Regulator

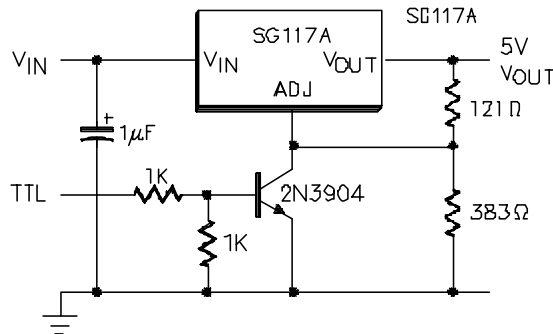
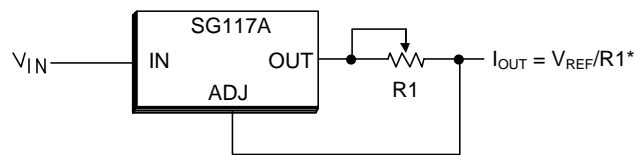


Figure 23 - 5V Regulator with Shut Down



* $0.8\Omega \leq R1 \leq 120\Omega$

Figure 24 -

Figure 25 - Programmable Current Limiter

PACKAGE OUTLINE DIMENSIONS

Controlling dimensions are in inches, metric equivalents are shown for general information.

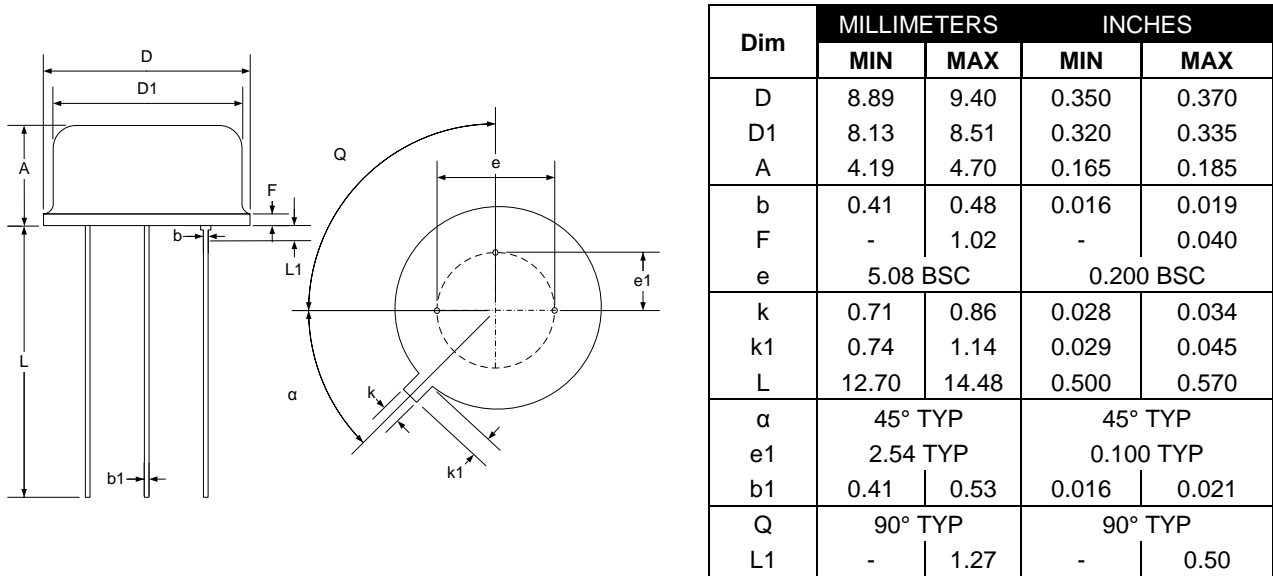
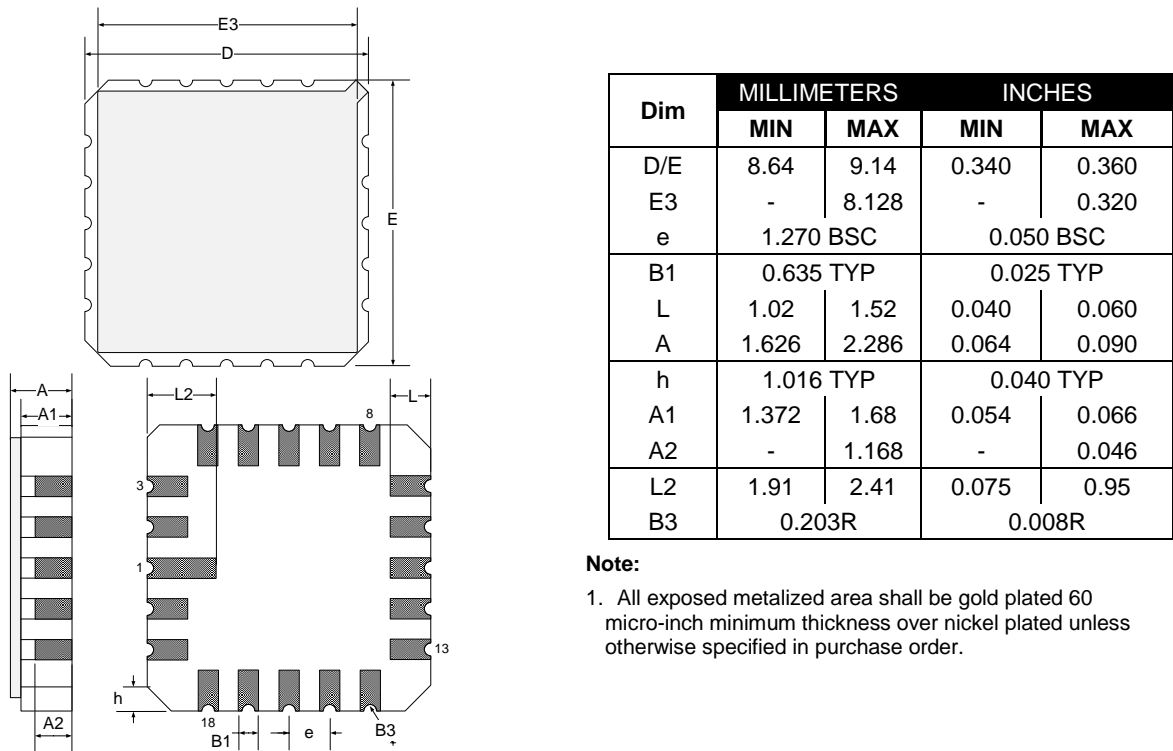


Figure 26 • T 3-Pin Metal Can TO-39 Package Dimensions

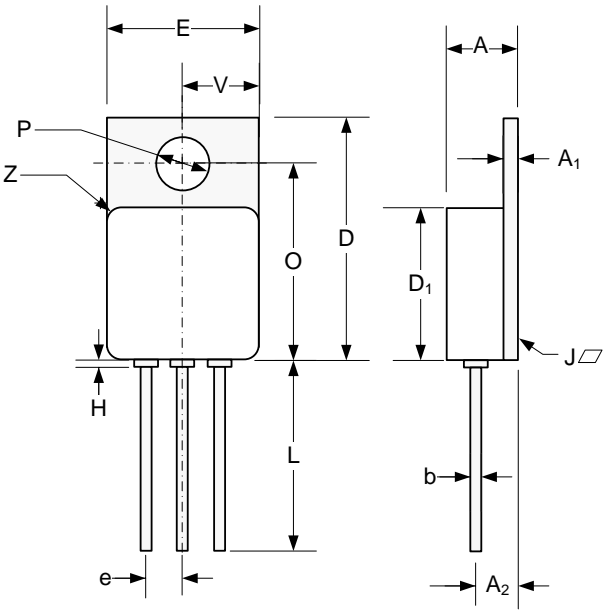


Note:

1. All exposed metalized area shall be gold plated 60 micro-inch minimum thickness over nickel plated unless otherwise specified in purchase order.

Figure 27 • L 20-Pin Ceramic Leadless Chip Carrier (LCC) Package Dimensions

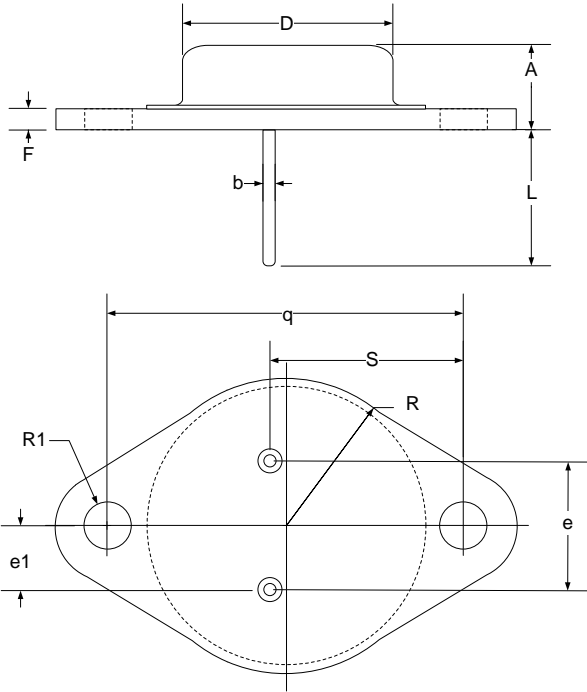
PACKAGE OUTLINE DIMENSIONS



Dim	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	4.70	5.21	0.185	0.205
A ₁	0.89	1.14	0.035	0.045
A ₂	2.92	3.18	0.115	0.125
b	0.71	.081	0.027	0.032
D	16.38	16.76	0.645	0.660
D ₁ *	10.41	10.92	0.410	0.430
e	2.54 BSC		0.100 BSC	
E*	10.41	10.67	0.410	0.420
H		0.50		0.020
L	12.70		0.500	
O	13.39	13.64	0.527	0.537
P	3.56	3.81	0.140	0.150
J		0.10		0.004
V	5.13	5.38	0.202	0.212
Z	1.40 TYP		0.055 TYP	

*Excludes Weld Fillet Around Lid.

Figure 28 • G/IG 3-Pin Hermetic TO-257 Package Dimensions



Dim	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	6.86	7.62	0.270	0.300
q	29.90	30.40	1.177	1.197
b	0.97	1.09	0.038	0.043
D	19.43	19.68	0.765	0.775
S	16.64	17.14	0.655	0.675
e	10.67	11.18	0.420	0.440
E1	5.21	5.72	0.205	0.225
F	1.52	2.03	0.060	0.080
R1	3.84	4.09	0.151	0.161
L	10.79	12.19	0.425	0.480
R	12.57	13.34	0.495	0.525

Figure 29 • K 3-Pin TO-3 Package Dimensions



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