



**THE DATASHEET OF  
LM2681M6X**



## LM2681 Switched Capacitor Voltage Converter

Check for Samples: [LM2681](#)

### FEATURES

- Doubles or Splits Input Supply Voltage
- SOT-23 6-Lead Package
- 15Ω Typical Output Impedance
- 90% Typical Conversion Efficiency at 20 mA

### APPLICATIONS

- Cellular Phones
- Pagers
- PDAs
- Operational Amplifier Power Suppliers
- Interface Power Suppliers
- Handheld Instruments

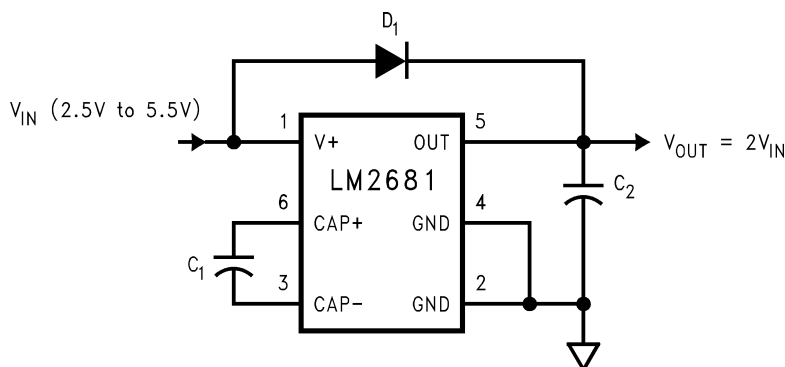
### DESCRIPTION

The LM2681 CMOS charge-pump voltage converter operates as a voltage doubler for an input voltage in the range of +2.5V to +5.5V. Two low cost capacitors and a diode (needed during start-up) is used in this circuit to provide up to 20 mA of output current. The LM2681 can also work as a voltage divider to split a voltage in the range of +1.8V to +11V in half.

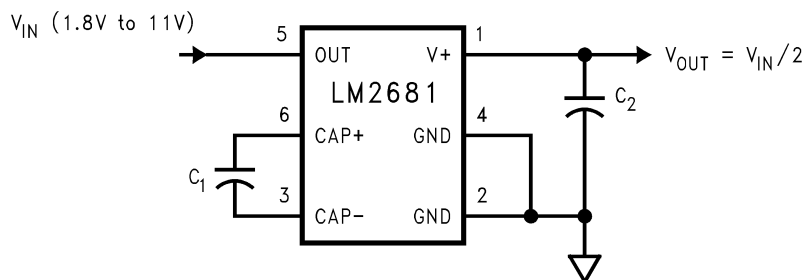
The LM2681 operates at 160 kHz oscillator frequency to reduce output resistance and voltage ripple. With an operating current of only 550 μA (operating efficiency greater than 90% with most loads) the LM2681 provides ideal performance for battery powered systems. The device is in a SOT-23, 6-lead package.

### Basic Application Circuits

#### Voltage Doubler



#### Splitting $V_{in}$ in Half



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## Absolute Maximum Ratings<sup>(1)(2)</sup>

V+ to GND Voltage	5.8V
OUT to GND Voltage	11.6V
OUT to V+ Voltage	5.8V
V+ and OUT Continuous Output Current	30 mA
Output Short-Circuit Duration to GND <sup>(3)</sup>	1 sec.
Continuous Power Dissipation (T <sub>A</sub> = 25°C) <sup>(4)</sup>	600 mW
T <sub>JMax</sub> <sup>(4)</sup>	150°C
θ <sub>JA</sub> <sup>(4)</sup>	210°C/W
Operating Junction Temperature Range	–40° to 85°C
Storage Temperature Range	–65°C to +150°C
Lead Temp. (Soldering, 10 seconds)	300°C
ESD Rating	2kV

- (1) Absolute maximum ratings indicate limits beyond which damage to the device may occur. Electrical specifications do not apply when operating the device beyond its rated operating conditions.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) OUT may be shorted to GND for one second without damage. However, shorting OUT to V+ may damage the device and should be avoided. Also, for temperatures above 85°C, OUT must not be shorted to GND or V+, or device may be damaged.
- (4) The maximum allowable power dissipation is calculated by using  $P_{DMax} = (T_{JMax} - T_A)/\theta_{JA}$ , where T<sub>JMax</sub> is the maximum junction temperature, T<sub>A</sub> is the ambient temperature, and θ<sub>JA</sub> is the junction-to-ambient thermal resistance of the specified package.

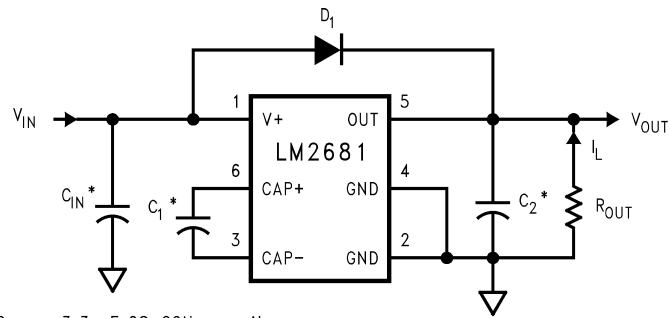
## Electrical Characteristics

Limits in standard typeface are for T<sub>J</sub> = 25°C, and limits in **boldface** type apply over the full operating temperature range. Unless otherwise specified: V+ = 5V, C<sub>1</sub> = C<sub>2</sub> = 3.3 μF.<sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
V+	Supply Voltage		<b>2.5</b>		<b>5.5</b>	V
I <sub>Q</sub>	Supply Current	No Load		550	<b>1000</b>	μA
I <sub>L</sub>	Output Current		<b>20</b>			mA
R <sub>SW</sub>	Sum of the R <sub>ds(on)</sub> of the four internal MOSFET switches	I <sub>L</sub> = 20 mA		8	<b>16</b>	Ω
R <sub>OUT</sub>	Output Resistance <sup>(2)</sup>	I <sub>L</sub> = 20 mA		15	<b>40</b>	Ω
f <sub>OSC</sub>	Oscillator Frequency	See <sup>(3)</sup>	<b>80</b>	160		kHz
f <sub>SW</sub>	Switching Frequency	See <sup>(3)</sup>	<b>40</b>	80		kHz
P <sub>EFF</sub>	Power Efficiency	R <sub>L</sub> (1.0k) between GND and OUT	<b>86</b>	93		%
		I <sub>L</sub> = 20 mA to GND		90		
V <sub>OEFF</sub>	Voltage Conversion Efficiency	No Load	<b>99</b>	99.96		%

- (1) In the test circuit, capacitors C<sub>1</sub> and C<sub>2</sub> are 3.3 μF, 0.3Ω maximum ESR capacitors. Capacitors with higher ESR will increase output resistance, reduce output voltage and efficiency.
- (2) Specified output resistance includes internal switch resistance and capacitor ESR. See [POSITIVE VOLTAGE DOUBLER](#)
- (3) The output switches operate at one half of the oscillator frequency, f<sub>OSC</sub> = 2f<sub>SW</sub>.

Test Circuit



\*  $C_{IN}$ ,  $C_1$ , and  $C_2$  are 3.3  $\mu\text{F}$  OS-CON capacitors.

Figure 1. LM2681 Test Circuit

### Typical Performance Characteristics

(Circuit of Figure 1, V+ = 5V unless otherwise specified)

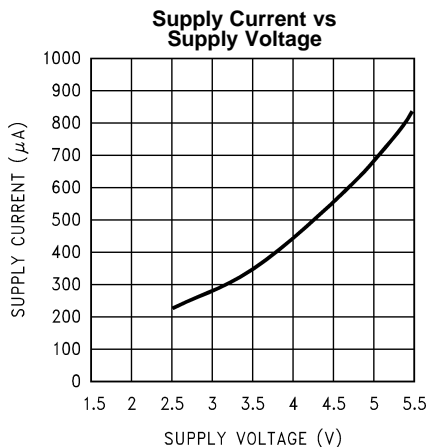


Figure 2.

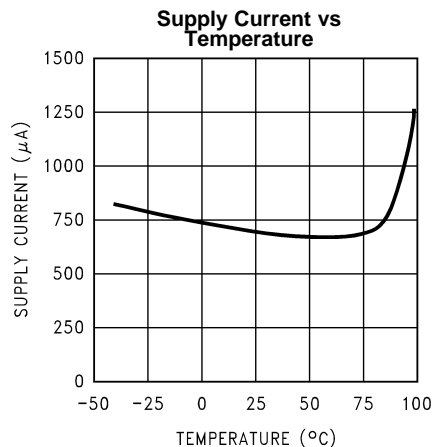


Figure 3.

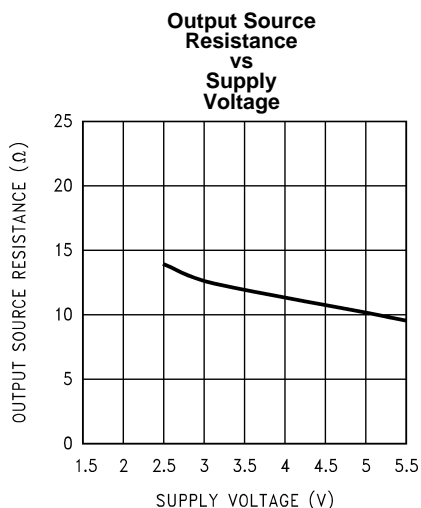


Figure 4.

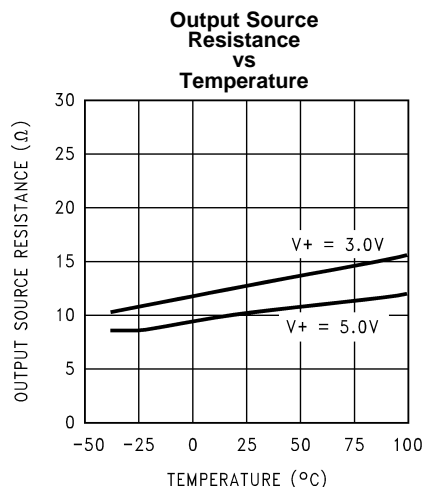


Figure 5.

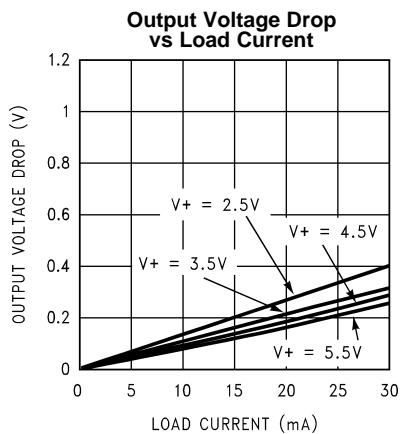


Figure 6.

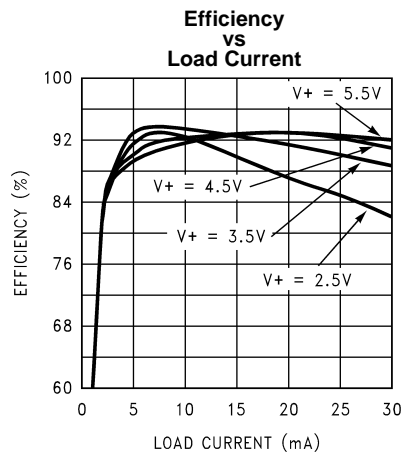
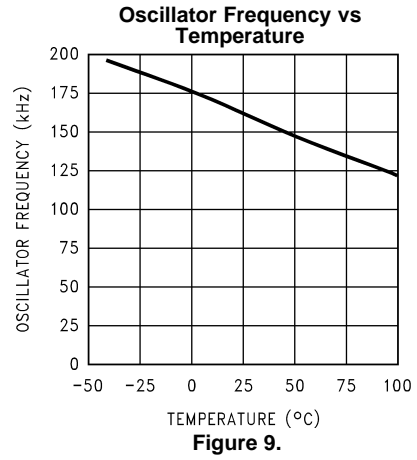
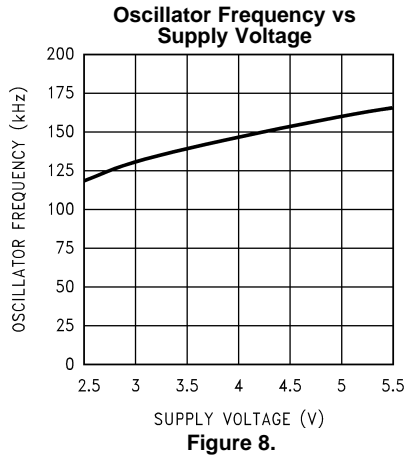


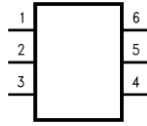
Figure 7.

**Typical Performance Characteristics (continued)**

(Circuit of [Figure 1](#),  $V+$  = 5V unless otherwise specified)



## CONNECTION DIAGRAM



**Figure 10. SOT-23, 6-Lead Package – Top View**  
See Package Number DBV



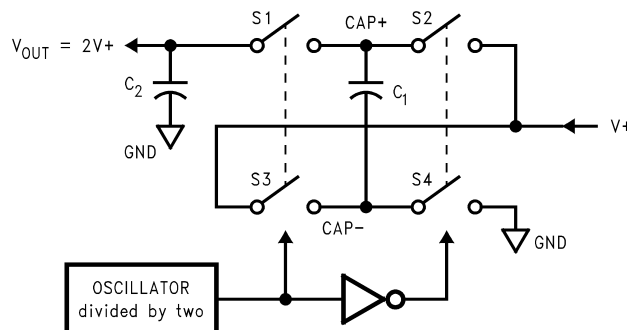
**Figure 11. Actual Size**

## PIN DESCRIPTION

Pin	Name	Function	
		Voltage Doubler	Voltage Split
1	V+	Power supply positive voltage input	Positive voltage output
2	GND	Power supply ground input	Same as doubler
3	CAP-	Connect this pin to the negative terminal of the charge-pump capacitor	Same as doubler
4	GND	Power supply ground input	Same as doubler
5	OUT	Positive voltage output	Power supply positive voltage input
6	CAP+	Connect this pin to the positive terminal of the charge-pump capacitor	Same as doubler

## Circuit Description

The LM2681 contains four large CMOS switches which are switched in a sequence to double the input supply voltage. Energy transfer and storage are provided by external capacitors. [Figure 12](#) illustrates the voltage conversion scheme. When  $S_2$  and  $S_4$  are closed,  $C_1$  charges to the supply voltage  $V_+$ . During this time interval, switches  $S_1$  and  $S_3$  are open. In the next time interval,  $S_2$  and  $S_4$  are open; at the same time,  $S_1$  and  $S_3$  are closed, the sum of the input voltage  $V_+$  and the voltage across  $C_1$  gives the  $2V_+$  output voltage when there is no load. The output voltage drop when a load is added is determined by the parasitic resistance ( $R_{ds(on)}$ ) of the MOSFET switches and the ESR of the capacitors and the charge transfer loss between capacitors. Details will be discussed in the [APPLICATION INFORMATION](#) section.



**Figure 12. Voltage Doubling Principle**

## APPLICATION INFORMATION

### POSITIVE VOLTAGE DOUBLER

The main application of the LM2681 is to double the input voltage. The range of the input supply voltage is 2.5V to 5.5V.

The output characteristics of this circuit can be approximated by an ideal voltage source in series with a resistance. The voltage source equals  $2V_+$ . The output resistance  $R_{out}$  is a function of the ON resistance of the internal MOSFET switches, the oscillator frequency, the capacitance and ESR of  $C_1$  and  $C_2$ . Since the switching current charging and discharging  $C_1$  is approximately twice as the output current, the effect of the ESR of the pumping capacitor  $C_1$  will be multiplied by four in the output resistance. The output capacitor  $C_2$  is charging and discharging at a current approximately equal to the output current, therefore, its ESR only counts once in the output resistance. A good approximation of  $R_{out}$  is:

$$R_{OUT} \cong 2R_{SW} + \frac{2}{f_{OSC} \times C_1} + 4ESR_{C1} + ESR_{C2} \quad (1)$$

where  $R_{SW}$  is the sum of the ON resistance of the internal MOSFET switches shown in [Figure 12](#).

The peak-to-peak output voltage ripple is determined by the oscillator frequency, the capacitance and ESR of the output capacitor  $C_2$ :

$$V_{RIPPLE} = \frac{I_L}{f_{OSC} \times C_2} + 2 \times I_L \times ESR_{C2} \quad (2)$$

High capacitance, low ESR capacitors can reduce both the output resistance and the voltage ripple.

The Schottky diode  $D_1$  is only needed for start-up. The internal oscillator circuit uses the OUT pin and the GND pin. Voltage across OUT and GND must be larger than 1.8V to insure the operation of the oscillator. During start-up,  $D_1$  is used to charge up the voltage at the OUT pin to start the oscillator; also, it protects the device from turning-on its own parasitic diode and potentially latching-up. Therefore, the Schottky diode  $D_1$  should have enough current carrying capability to charge the output capacitor at start-up, as well as a low forward voltage to prevent the internal parasitic diode from turning-on. A Schottky diode like 1N5817 can be used for most applications. If the input voltage ramp is less than 10V/ms, a smaller Schottky diode like MBR0520LT1 can be used to reduce the circuit size.

### SPLIT V+ IN HALF

Another interesting application shown in the Basic Application Circuits is using the LM2681 as a precision voltage divider. This circuit can be derived from the voltage doubler by switching the input and output connections. In the voltage divider, the input voltage applies across the OUT pin and the GND pin (which are the power rails for the internal oscillator), therefore no start-up diode is needed. Also, since the off-voltage across each switch equals  $V_{in}/2$ , the input voltage can be raised to +11V.

### CAPACITOR SELECTION

As discussed in the [POSITIVE VOLTAGE DOUBLER](#) section, the output resistance and ripple voltage are dependent on the capacitance and ESR values of the external capacitors. The output voltage drop is the load current times the output resistance, and the power efficiency is

$$\eta = \frac{P_{OUT}}{P_{IN}} = \frac{I_L^2 R_L}{I_L^2 R_L + I_L^2 R_{OUT} + I_Q(V_+)} \quad (3)$$

Where  $I_Q(V_+)$  is the quiescent power loss of the IC device, and  $I_L^2 R_{out}$  is the conversion loss associated with the switch on-resistance, the two external capacitors and their ESRs.

The selection of capacitors is based on the specifications of the dropout voltage (which equals  $I_{out} R_{out}$ ), the output voltage ripple, and the converter efficiency. Low ESR capacitors are recommended to maximize efficiency, reduce the output voltage drop and voltage ripple.

**Low ESR Capacitor Manufacturers**

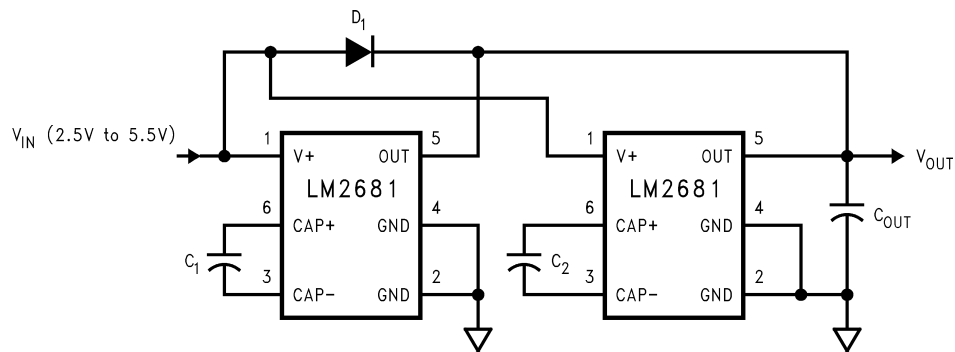
Manufacturer	Phone	Capacitor Type
Nichicon Corp.	(708)-843-7500	PL & PF series, through-hole aluminum electrolytic
AVX Corp.	(803)-448-9411	TPS series, surface-mount tantalum
Sprague	(207)-324-4140	593D, 594D, 595D series, surface-mount tantalum
Sanyo	(619)-661-6835	OS-CON series, through-hole aluminum electrolytic
Murata	(800)-831-9172	Ceramic chip capacitors
Taiyo Yuden	(800)-348-2496	Ceramic chip capacitors
Token	(408)-432-8020	Ceramic chip capacitors

**Other Applications**

**PARALLELING DEVICES**

Any number of LM2681s can be paralleled to reduce the output resistance. Each device must have its own pumping capacitor  $C_1$ , while only one output capacitor  $C_{out}$  is needed as shown in Figure 13. The composite output resistance is:

$$R_{OUT} = \frac{R_{OUT} \text{ of each LM2681}}{\text{Number of Devices}} \tag{4}$$



**Figure 13. Lowering Output Resistance by Paralleling Devices**

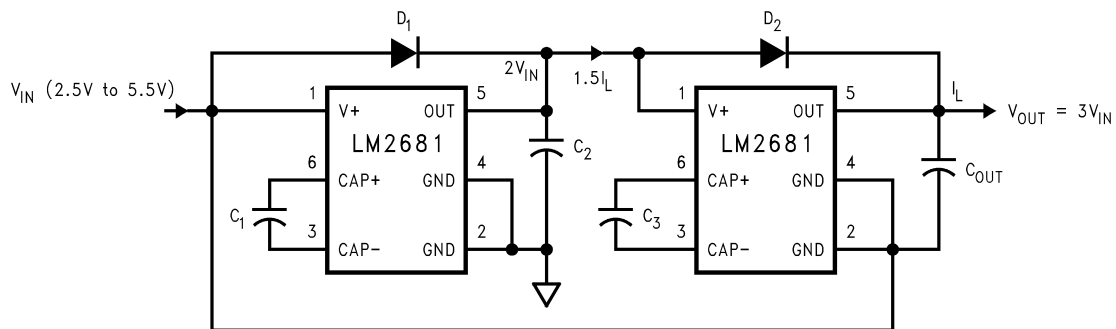
**CASCADING DEVICES**

Cascading the LM2681s is an easy way to produce a greater voltage (A two-stage cascade circuit is shown in Figure 14).

The effective output resistance is equal to the weighted sum of each individual device:

$$R_{out} = 1.5R_{out\_1} + R_{out\_2} \tag{5}$$

Note that, the increasing of the number of cascading stages is practically limited since it significantly reduces the efficiency, increases the output resistance and output voltage ripple.



**Figure 14. Increasing Output Voltage by Cascading Devices**

## REGULATING $V_{OUT}$

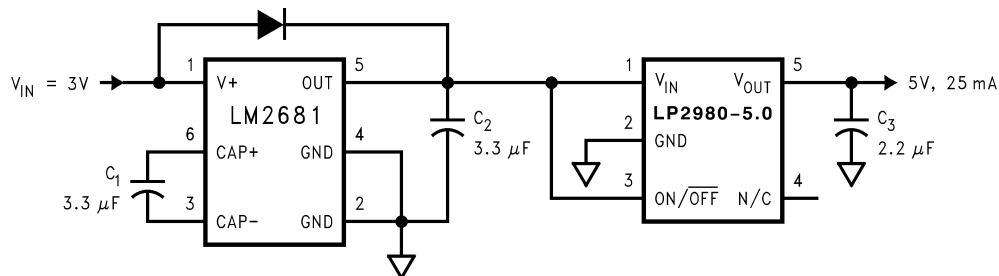
It is possible to regulate the output of the LM2681 by use of a low dropout regulator (such as LP2980-5.0). The whole converter is depicted in [Figure 15](#).

A different output voltage is possible by use of LP2980-3.3, LP2980-3.0, or LP2980-adj.

Note that, the following conditions must be satisfied simultaneously for worst case design:

$$2V_{in\_min} > V_{out\_min} + V_{drop\_max} (LP2980) + I_{out\_max} \times R_{out\_max} (LM2681) \quad (6)$$

$$2V_{in\_max} < V_{out\_max} + V_{drop\_min} (LP2980) + I_{out\_min} \times R_{out\_min} (LM2681) \quad (7)$$



**Figure 15. Generate a Regulated +5V from +3V Input Voltage**

### REVISION HISTORY

Changes from Revision A (May 2013) to Revision B	Page
• Changed layout of National Data Sheet to TI format .....	<a href="#">9</a>

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM2681M6/NOPB	ACTIVE	SOT-23	DBV	6	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	S10A	<a href="#">Samples</a>
LM2681M6X/NOPB	ACTIVE	SOT-23	DBV	6	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	S10A	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2681M6/NOPB	SOT-23	DBV	6	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LM2681M6X/NOPB	SOT-23	DBV	6	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

**TAPE AND REEL BOX DIMENSIONS**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2681M6/NOPB	SOT-23	DBV	6	1000	210.0	185.0	35.0
LM2681M6X/NOPB	SOT-23	DBV	6	3000	210.0	185.0	35.0

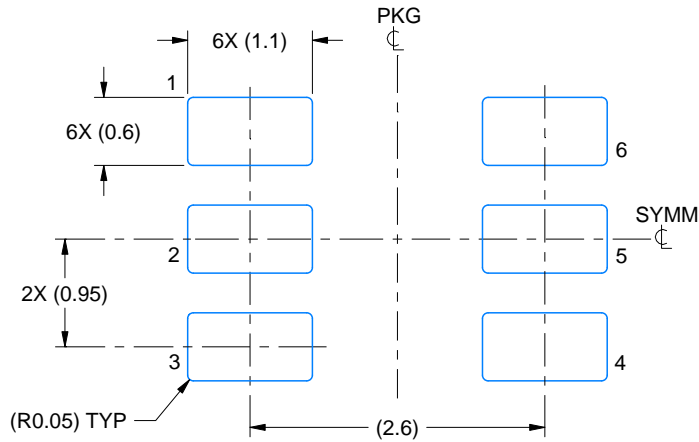


# EXAMPLE BOARD LAYOUT

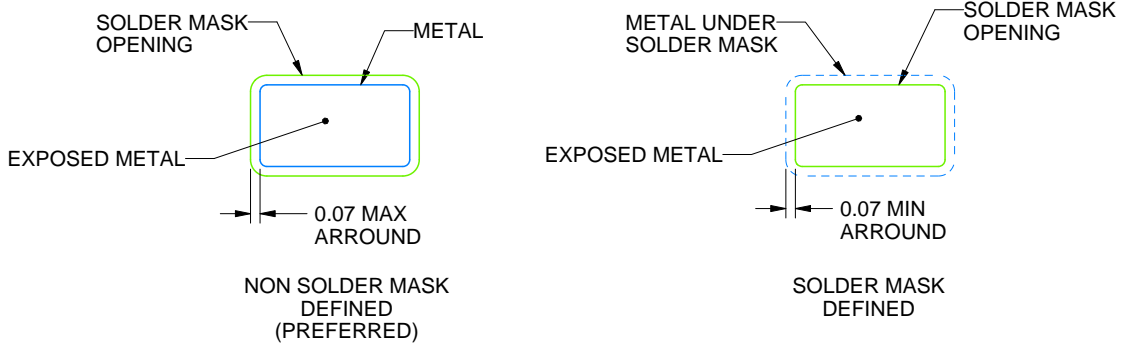
DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214840/B 03/2018

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

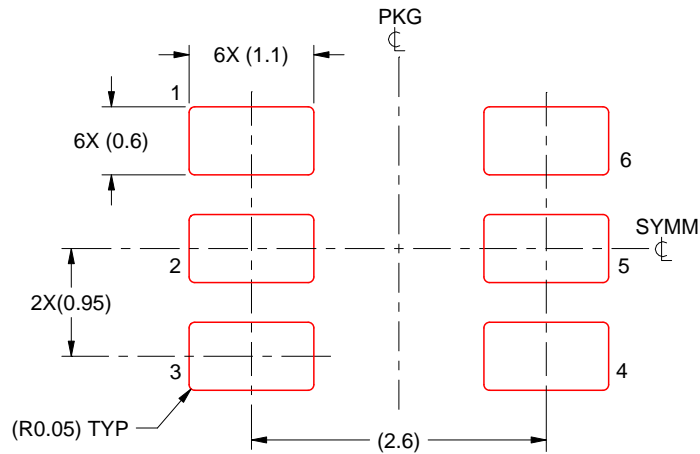
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214840/B 03/2018

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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