



**THE DATASHEET OF  
NCV4299D2G**



# NCV4299

## 150 mA Low-Dropout Voltage Regulator

The NCV4299 is a family of precision micropower voltage regulators with an output current capability of 150 mA. It is available in 5.0 V or 3.3 V output voltage, and is housed in an 8-lead SOIC and in a 14-lead SOIC (fused) package.

The output voltage is accurate within  $\pm 2\%$  with a maximum dropout voltage of 0.5 V at 100 mA. Low Quiescent current is a feature drawing only 90  $\mu\text{A}$  with a 1 mA load. This part is ideal for any and all battery operated microprocessor equipment.

The device features microprocessor interfaces including an adjustable reset output and adjustable system monitor to provide shutdown early warning. An inhibit function is available on the 14-lead part. With inhibit active, the regulator turns off and the device consumes less than 1.0  $\mu\text{A}$  of quiescent current.

The part can withstand load dump transients making it suitable for use in automotive environments.

### Features

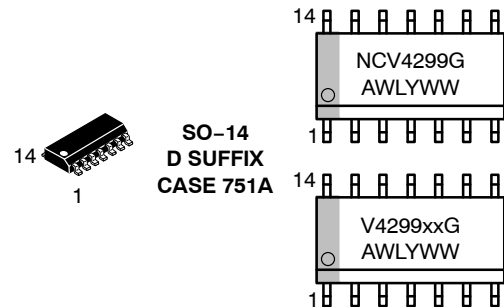
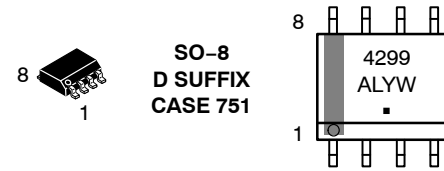
- 5.0 V, 3.3 V  $\pm 2\%$ , 150 mA
- Extremely Low Current Consumption
  - ◆ 90  $\mu\text{A}$  (Typ) in the ON Mode
  - ◆ < 1.0  $\mu\text{A}$  in the Off Mode
- Early Warning
- Reset Output Low Down to  $V_Q = 1.0\text{ V}$
- Adjustable Reset Threshold
- Wide Temperature Range
- Fault Protection
  - ◆ 60 V Peak Transient Voltage
  - ◆ -40 V Reverse Voltage
  - ◆ Short Circuit
  - ◆ Thermal Overload
- Internally Fused Leads in the SO-14 Package
- Inhibit Function with  $\mu\text{A}$  Current Consumption in the Off Mode
- NCV Prefix for Automotive and Other Applications Requiring Site and Change Control
- These are Pb-Free Devices



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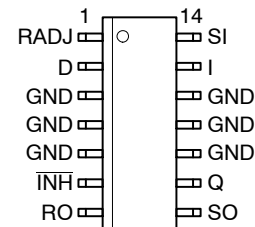
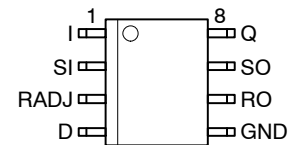
<http://onsemi.com>

### MARKING DIAGRAMS



xx = 33 (3.3 V Version)  
 A = Assembly Location  
 L, WL = Wafer Lot  
 Y = Year  
 W, WW = Work Week  
 G or ■ = Pb-Free Package

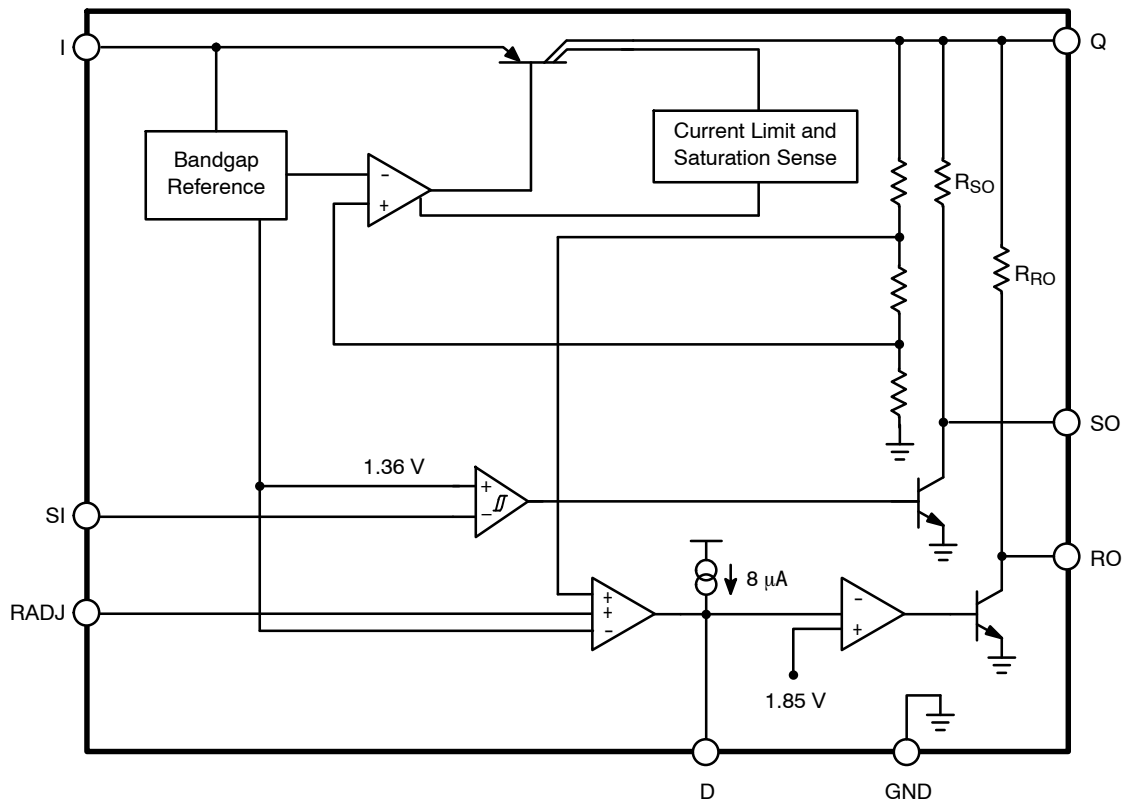
### PIN CONNECTIONS



### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 21 of this data sheet.

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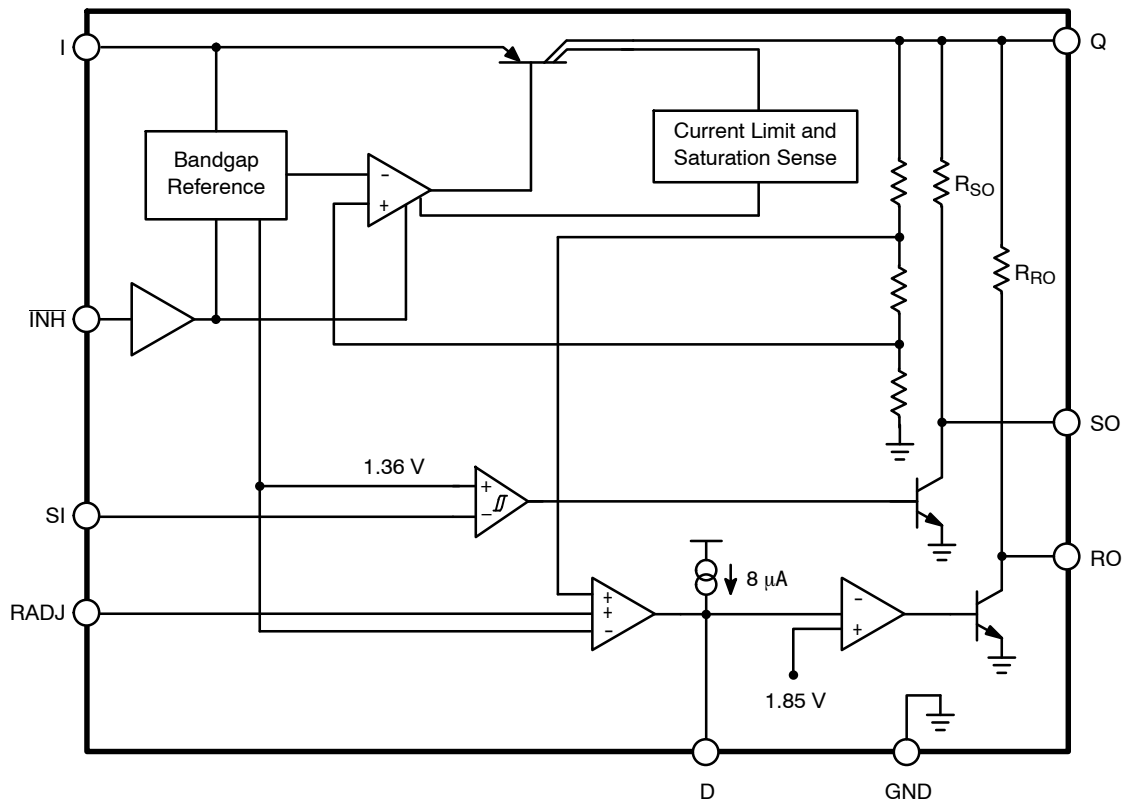


**Figure 1. SO-8 Simplified Block Diagram**

## PIN FUNCTION DESCRIPTION – SO-8 PACKAGE

Pin	Symbol	Description
1	I	Input. Battery Supply Input Voltage. Bypass directly to GND with ceramic capacitor.
2	SI	Sense Input. Can provide an early warning signal of an impending reset condition when used with SO. Connect to Q if not used.
3	RADJ	Reset Adjust. Use resistor divider to Q to adjust reset threshold lower. Connect to GND if not used.
4	D	Reset Delay. Connect external capacitor to ground to set delay time.
5	GND	Ground.
6	RO	Reset Output. NPN collector output with internal 20 kΩ pullup to Q. Notifies user of out of regulation condition. Leave open if not used.
7	SO	Sense Output. NPN collector output with internal 20 kΩ pullup to Q. Can be used to provide early warning of an impending reset condition. Leave open if not used.
8	Q	5.0 V, 3.3 V, ±2%, 150 mA output. Use 22 µF, ESR < 5.0 Ω to ground.

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**Figure 2. SO-14 Simplified Block Diagram**

## PIN FUNCTION DESCRIPTION – SO-14 PACKAGE

Pin	Symbol	Description
1	RADJ	Reset Adjust. Use resistor divider to Q to adjust reset threshold lower. Connect to GND if not used.
2	D	Reset Delay. Connect external capacitor to ground to set delay time.
3	GND	Ground.
4	GND	Ground.
5	GND	Ground.
6	INH	Inhibit. Connect to I if not needed. A high turns the regulator on. Use a low pass filter if transients with slew rate in excess of 10 V/ms may be present on this pin during operation. See Figure 40 for details.
7	RO	Reset Output. NPN collector output with internal 20 kΩ pullup to Q. Notifies user of out of regulation condition.
8	SO	Sense Output. NPN collector output with internal 20 kΩ pullup to Q. Can be used to provide early warning of an impending reset condition.
9	Q	5.0 V, 3.3 V, ±2%, 150 mA output. Use 22 µF, ESR < 5.0 Ω to ground.
10	GND	Ground.
11	GND	Ground.
12	GND	Ground.
13	I	Input. Battery Supply Input Voltage.
14	SI	Sense Input. Can provide an early warning signal of an impending reset condition when used with SO.

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## MAXIMUM RATINGS

Rating	Symbol	Min	Max	Unit
Input Voltage to Regulator (DC)	$V_I$	-40	45	V
Input Peak Transient Voltage to Regulator wrt GND	-	-	60	V
Inhibit ( $\overline{INH}$ ) (Note 1)	$V_{\overline{INH}}$	-40	45	V
Sense Input (SI)	$V_{SI}$	-0.3	45	V
Sense Input (SI)	$I_{SI}$	-1.0	1.0	mA
Reset Threshold (RADJ)	$V_{RADJ}$	-0.3	7.0	V
Reset Threshold (RADJ)	$I_{RADJ}$	-10	10	mA
Reset Delay (D)	$V_D$	-0.3	7.0	V
Reset Output (RO)	$V_{RO}$	-0.3	7.0	V
Sense Output (SO)	$V_{SO}$	-0.3	7.0	V
Output (Q)	$V_Q$	-0.3	16	V
Output (Q)	$I_Q$	-5.0	-	mA
ESD Capability, Human Body Model (Note 3)	$ESD_{HB}$	2.0	-	kV
ESD Capability, Machine Model (Note 3)	$ESD_{MM}$	200	-	V
ESD Capability, Charged Device Model (Note 3)	$ESD_{CDM}$	1.0	-	kV
Junction Temperature	$T_J$	-	150	°C
Storage Temperature	$T_{stg}$	-50	150	°C

## OPERATING RANGE

Input Voltage 5.0 V Version 3.3 V Version	$V_I$	4.5 4.4	45 45	V
Junction Temperature	$T_J$	-40	150	°C

## LEAD TEMPERATURE SOLDERING REFLOW (Note 2)

Reflow (SMD styles only), lead free 60s-150 sec above 217, 40 sec max at peak	$T_{SLD}$	-	265 Pk	°C
Moisture Sensitivity Level	MSL	Level 1		

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

- 14 pin package only.
- Per IPC / JEDEC J-STD-020C.
- This device series incorporates ESD protection and is tested by the following methods:  
ESD HBM tested per AEC-Q100-002 (EIA/JESD22-A114)  
ESD MM tested per AEC-Q100-003 (EIA/JESD22-A115)  
ESD CDM tested per EIA/JES D22/C101, Field Induced Charge Model.

## THERMAL CHARACTERISTICS

Characteristic	Test Conditions (Typical Value)			Unit
	Note 4	Note 5	Note 6	
SO-8 Junction-to-Tab ( $\psi_{JLx}$ , $\theta_{JLx}$ ) Junction-to-Ambient ( $R_{\theta JA}$ , $\theta_{JA}$ )	54 172	52 144	48 118	°C/W
SO-14 Junction-to-Tab ( $\psi_{JLx}$ , $\theta_{JLx}$ ) Junction-to-Ambient ( $R_{\theta JA}$ , $\theta_{JA}$ )	19 112	21 89	20 67	°C/W

- 2 oz Copper, 50 mm sq Copper area, 1.5 mm thick FR4
- 2 oz Copper, 150 mm sq Copper area, 1.5 mm thick FR4
- 2 oz Copper, 500 mm sq Copper area, 1.5 mm thick FR4

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## ELECTRICAL CHARACTERISTICS (-40°C < T<sub>J</sub> < 150°C; V<sub>I</sub> = 13.5 V unless otherwise noted.)

Characteristic	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>Output Q</b>						
Output Voltage (5.0 V Version)	V <sub>Q</sub>	1.0 mA < I <sub>Q</sub> < 150 mA, 6.0 V < V <sub>I</sub> < 16 V	4.9	5.0	5.1	V
Output Voltage (3.3 V Version)	V <sub>Q</sub>	1.0 mA < I <sub>Q</sub> < 150 mA, 5.5 V < V <sub>I</sub> < 16 V	3.23	3.3	3.37	V
Current Limit	I <sub>Q</sub>	-	250	400	500	mA
Quiescent Current (I <sub>q</sub> = I <sub>I</sub> - I <sub>Q</sub> )	I <sub>q</sub>	INH ON, I <sub>Q</sub> < 1.0 mA, T <sub>J</sub> = 25°C	-	86	100	μA
Quiescent Current (I <sub>q</sub> = I <sub>I</sub> - I <sub>Q</sub> )	I <sub>q</sub>	INH ON, I <sub>Q</sub> < 1.0 mA	-	90	105	μA
Quiescent Current (I <sub>q</sub> = I <sub>I</sub> - I <sub>Q</sub> )	I <sub>q</sub>	INH ON, I <sub>Q</sub> = 10 mA	-	170	500	μA
Quiescent Current (I <sub>q</sub> = I <sub>I</sub> - I <sub>Q</sub> )	I <sub>q</sub>	INH ON, I <sub>Q</sub> = 50 mA	-	0.7	2.0	mA
Quiescent Current (I <sub>q</sub> = I <sub>I</sub> - I <sub>Q</sub> )	I <sub>q</sub>	INH = 0 V, T <sub>J</sub> = 25°C	-	-	1.0	μA
Dropout Voltage (Note 7)	V <sub>dr</sub>	I <sub>Q</sub> = 100 mA	-	0.22	0.50	V
Load Regulation	ΔV <sub>Q</sub>	I <sub>Q</sub> = 1.0 mA to 100 mA	-	5.0	30	mV
Line Regulation	ΔV <sub>Q</sub>	V <sub>I</sub> = 6.0 V to 28 V, I <sub>Q</sub> = 1.0 mA	-	10	25	mV
Power Supply Ripple Rejection	PSRR	f <sub>r</sub> = 100 Hz, V <sub>r</sub> = 1.0 V <sub>pp</sub> , I <sub>Q</sub> = 100 mA	-	66	-	dB

### Inhibit (INH) (14 Pin Package Only)

Inhibit Off Voltage	V <sub>INHOFF</sub>	V <sub>Q</sub> < 1.0 V	-	-	0.8	V
Inhibit On Voltage 5.0 V Version 3.3 V Version	V <sub>INHON</sub>	V <sub>Q</sub> > 4.85 V V <sub>Q</sub> > 3.2 V	3.5 3.5	- -	- -	V
Input Current	I <sub>INHON</sub> I <sub>INHOFF</sub>	INH ON INH = 0 V	- -	3.0 0.5	10 2.0	μA

### Reset (RO)

Switching Threshold 5.0 V Version 3.3 V Version	V <sub>RT</sub>	-	4.50 2.96	4.64 3.04	4.80 3.16	V
Output Resistance	R <sub>RO</sub>	-	10	20	40	kΩ
Reset Output Low Voltage 5.0 V Version 3.3 V Version	V <sub>RO</sub>	Q < 4.5 V, Internal R <sub>RO</sub> , I <sub>RO</sub> = -1.0 mA Q < 2.96 V, Internal R <sub>RO</sub> , I <sub>RO</sub> = -1.0 mA	- -	0.17 0.17	0.40 0.40	V
Allowable External Reset Pullup Resistor	V <sub>ROext</sub>	External Resistor to Q	5.6	-	-	kΩ
Delay Upper Threshold	V <sub>UD</sub>	-	1.5	1.85	2.2	V
Delay Lower Threshold	V <sub>LD</sub>	-	0.4	0.5	0.6	V

7. Measured when the output voltage V<sub>Q</sub> has dropped 100 mV from the nominal value obtained at V<sub>I</sub> = 13.5 V.

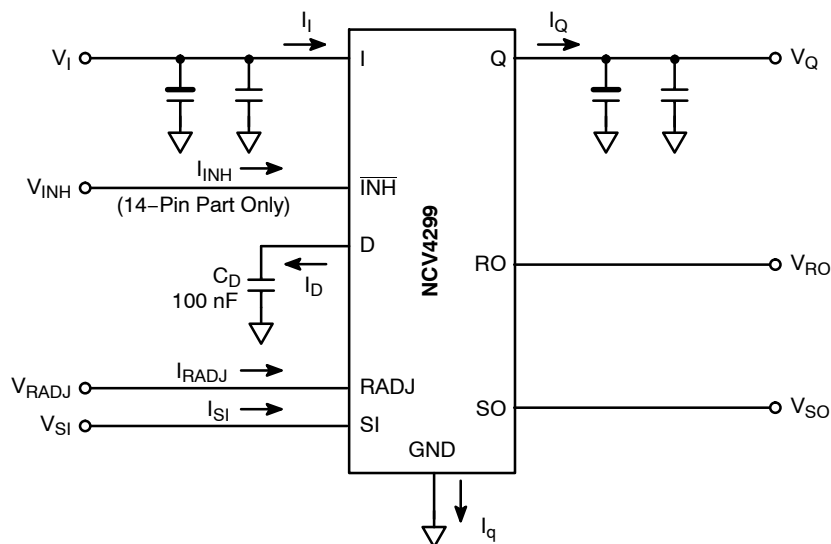
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**ELECTRICAL CHARACTERISTICS (continued)** ( $-40^{\circ}\text{C} < T_J < 150^{\circ}\text{C}$ ;  $V_I = 13.5\text{ V}$  unless otherwise noted.)

Characteristic	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>Reset (RO)</b>						
Delay Output Low Voltage 5.0 V Version 3.3 V Version	$V_{D,sat}$	$Q < 4.5\text{ V}$ , Internal $R_{RO}$ $Q < 2.96\text{ V}$ , Internal $R_{RO}$	- -	- 0.017	0.1 0.1	V
Delay Charge Current 5.0 V Version 3.3 V Version	$I_D$	$Q < 4.5\text{ V}$ , Internal $R_{RO}$ , $V_D = 1.0\text{ V}$ $Q < 2.96\text{ V}$ , Internal $R_{RO}$ , $V_D = 1.0\text{ V}$	4.0 -	7.1 -	12 -	$\mu\text{A}$
Power On Reset Delay Time	$t_d$	$C_D = 100\text{ nF}$	17	28	35	ms
Reset Reaction Time	$t_{RR}$	$C_D = 100\text{ nF}$	0.5	2.2	4.0	$\mu\text{s}$
Reset Adjust Switching Threshold 5.0 V Version 3.3 V Version	$V_{RADJ,TH}$	$Q > 3.5\text{ V}$ $Q > 2.3\text{ V}$	1.26 -	1.36 -	1.44 -	V

**Input Voltage Sense (SI and SO)**

Sense Input Threshold High	$V_{SI,High}$	-	1.34	1.45	1.54	V
Sense Input Threshold Low	$V_{SI,Low}$	-	1.26	1.36	1.44	V
Sense Input Hysteresis	-	(Sense Threshold High) - (Sense Threshold Low)	50	90	130	mV
Sense Input Current	$I_{SI}$	-	-1.0	0.1	1.0	$\mu\text{A}$
Sense Output Resistance	$R_{SO}$	-	10	20	40	$\text{k}\Omega$
Sense Output Low Voltage	$V_{SO}$	$V_{SI} < 1.20\text{ V}$ , $V_I > 4.2\text{ V}$ , $I_{SO} = 0\text{ }\mu\text{A}$	-	0.1	0.4	V
Allowable External Sense Out Pullup Resistor	$R_{SOext}$	-	5.6	-	-	$\text{k}\Omega$
SI High to SO High Reaction Time	$t_{PSOLH}$	-	-	4.4	8.0	$\mu\text{s}$
SI Low to SO Low Reaction Time	$t_{PSOHL}$	-	-	3.8	5.0	$\mu\text{s}$



**Figure 3. Measurement Circuit**

TYPICAL PERFORMANCE CHARACTERISTICS – 5.0 V OPTION

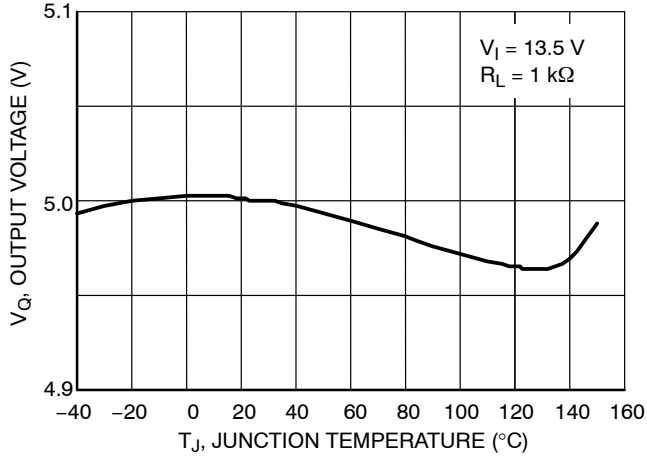


Figure 4. Output Voltage vs. Junction Temperature

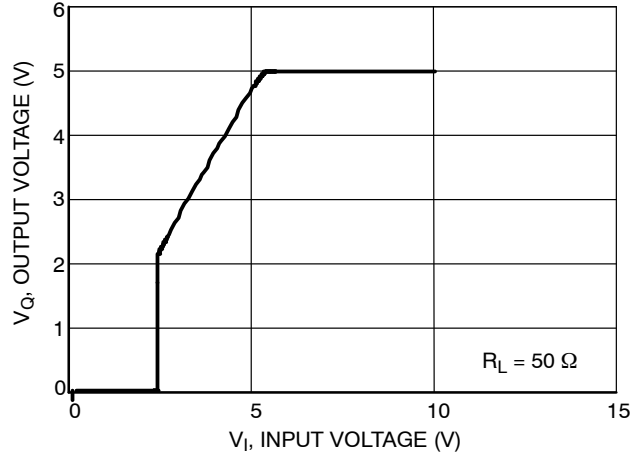


Figure 5. Output Voltage vs. Input Voltage

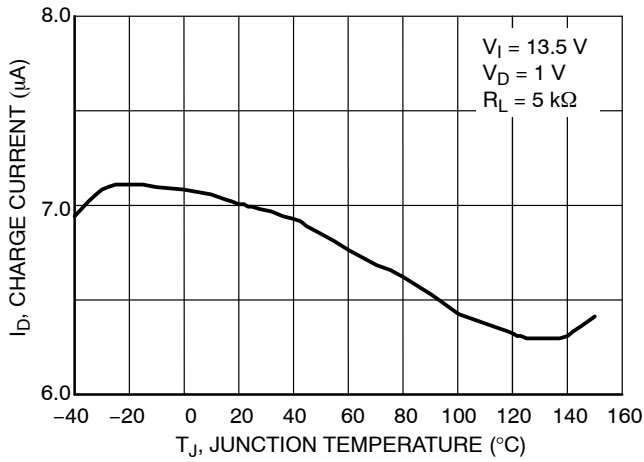


Figure 6. Charge Current vs. Junction Temperature

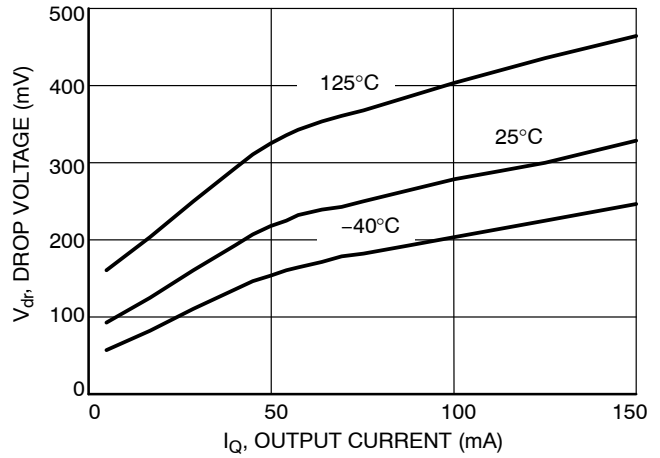


Figure 7. Drop Voltage vs. Output Current

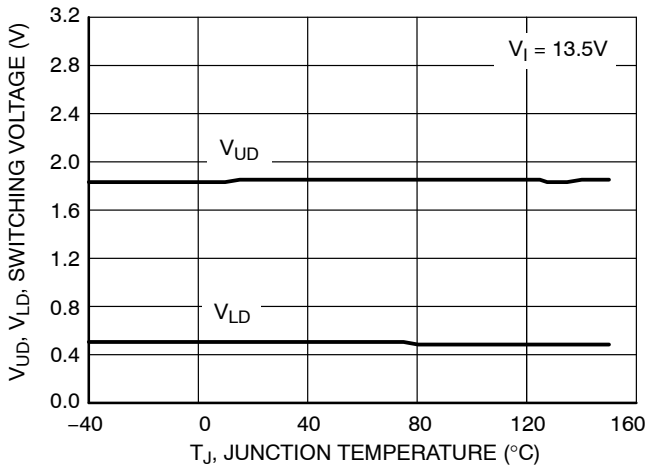


Figure 8. Switching Voltage vs. Junction Temperature

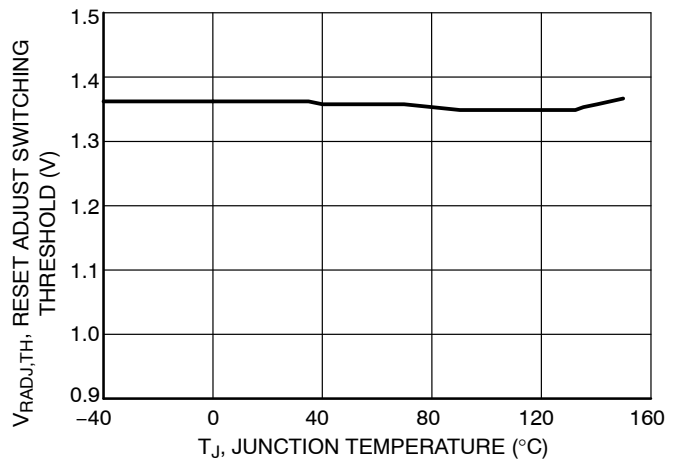


Figure 9. Reset Adjust Switching Threshold vs. Junction Temperature

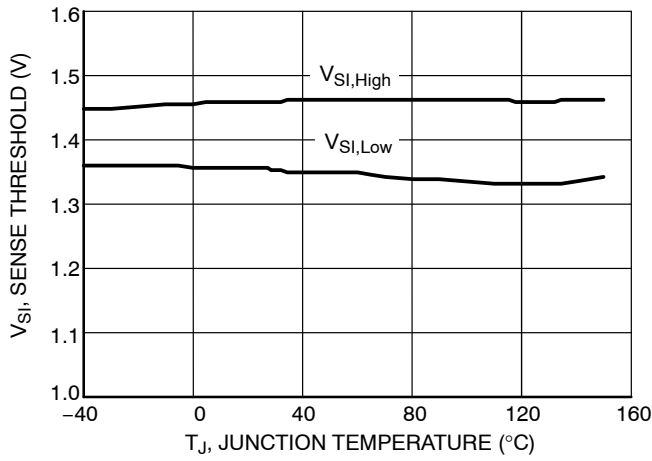


Figure 10. Sense Threshold vs. Junction Temperature

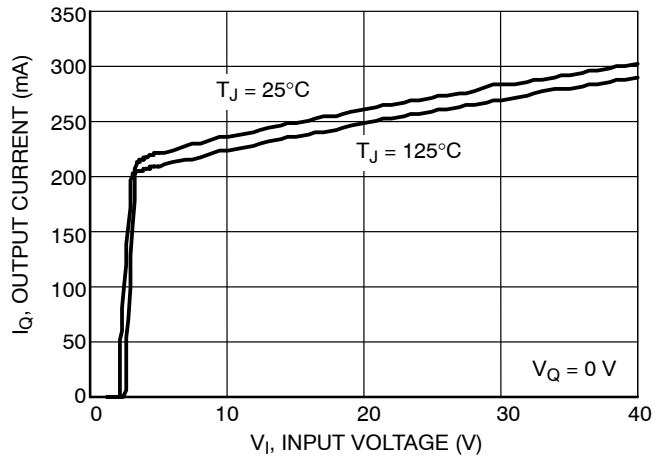


Figure 11. Output Current Limit vs. Input Voltage

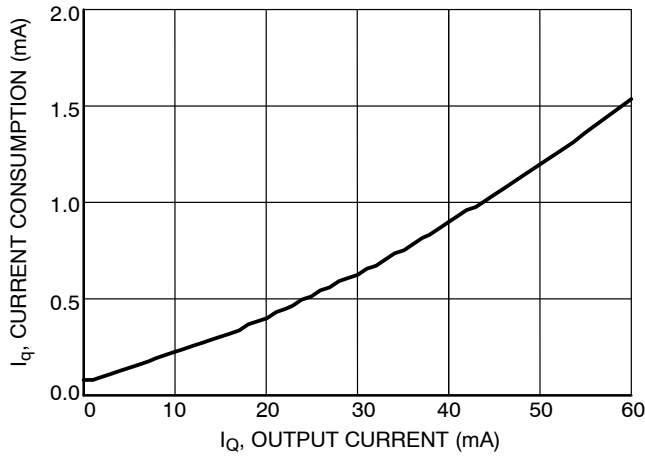


Figure 12. Current Consumption vs. Output Current

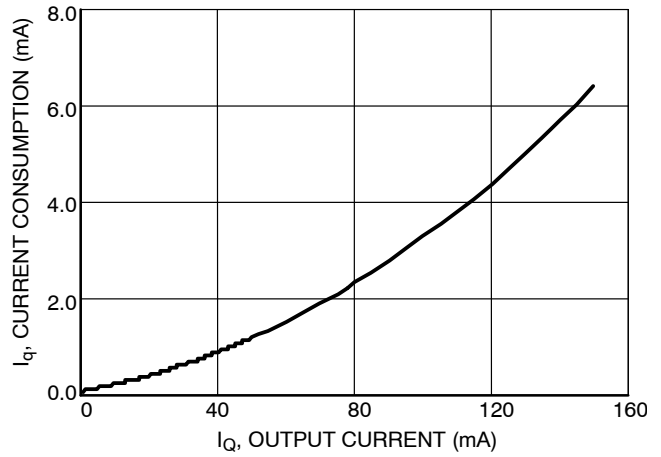


Figure 13. Current Consumption vs. Output Current

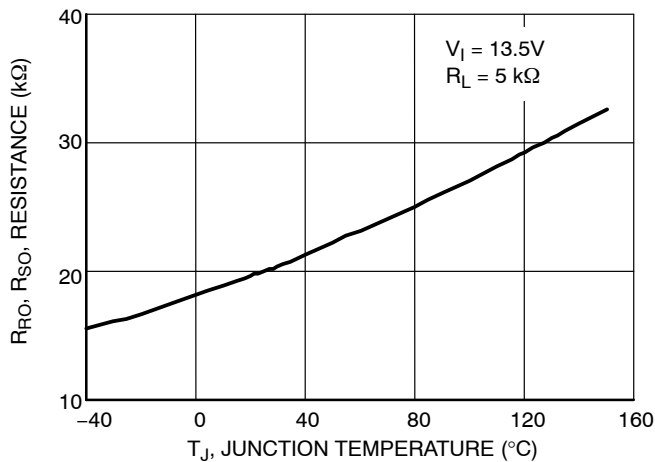


Figure 14.  $R_{RO}$ ,  $R_{SO}$  Resistance vs. Junction Temperature

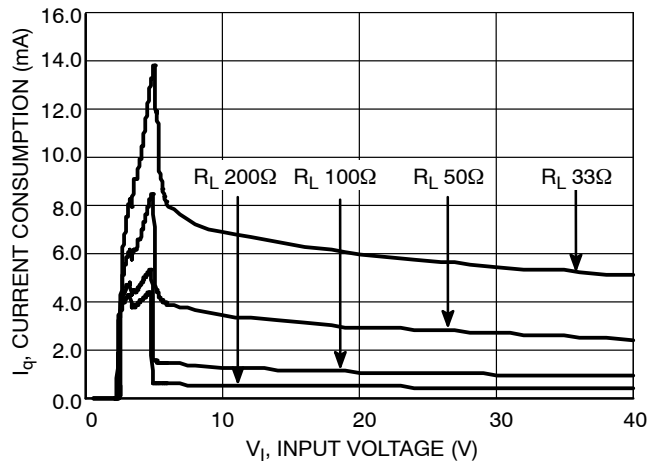
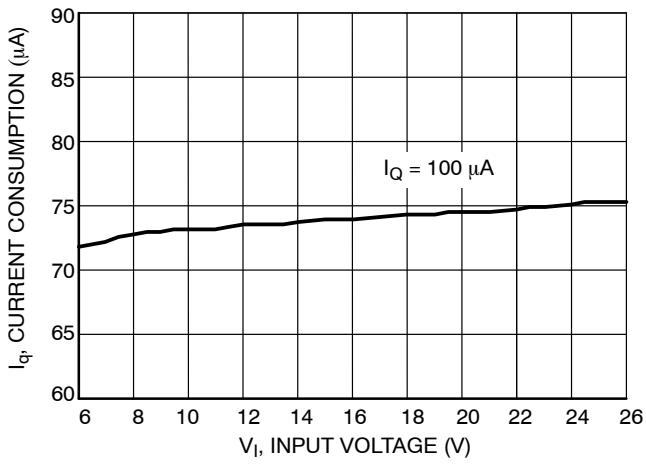
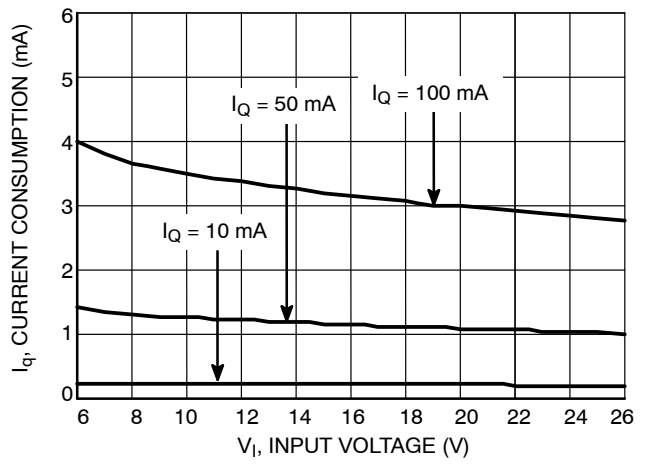


Figure 15. Current Consumption vs. Input Voltage

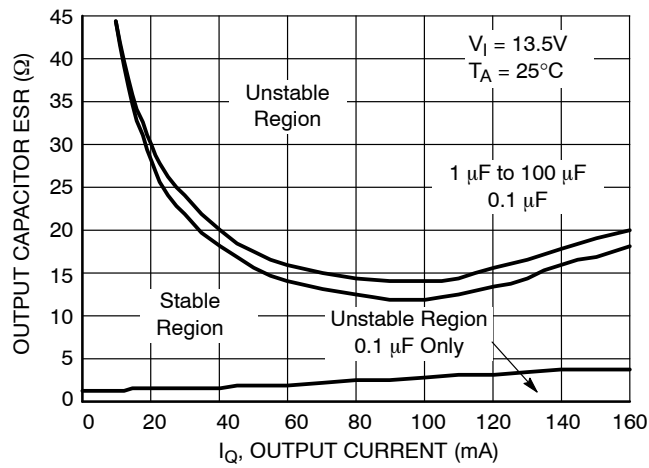
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**Figure 16. Current Consumption vs. Input Voltage**



**Figure 17. Current Consumption vs. Input Voltage**



**Figure 18. Output Stability vs. Output Capacitor ESR**

TYPICAL PERFORMANCE CHARACTERISTICS – 3.3 V OPTION

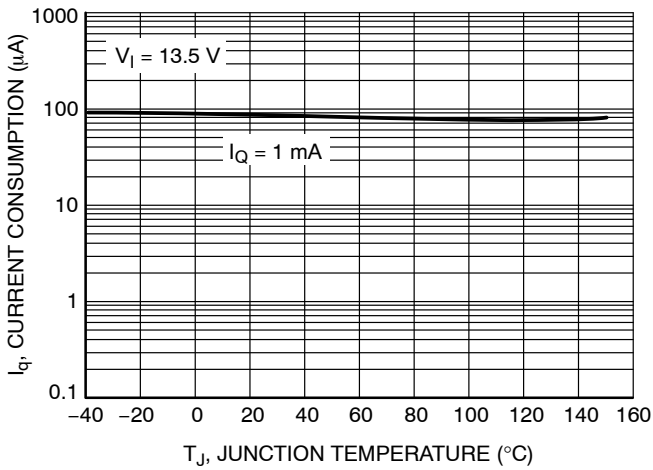


Figure 19. Current Consumption vs. Junction Temperature

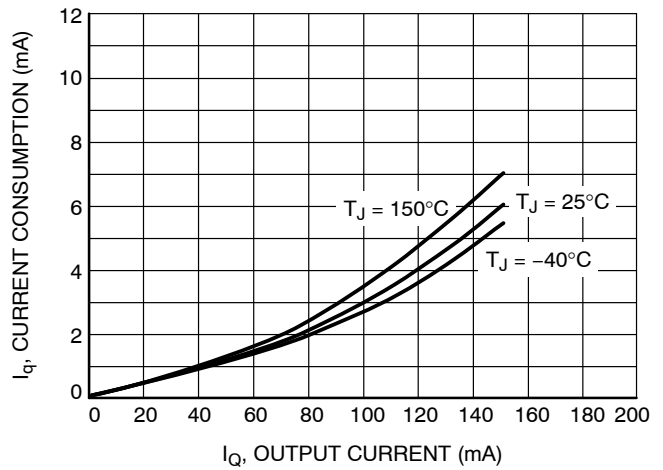


Figure 20. Current Consumption vs. Output Current

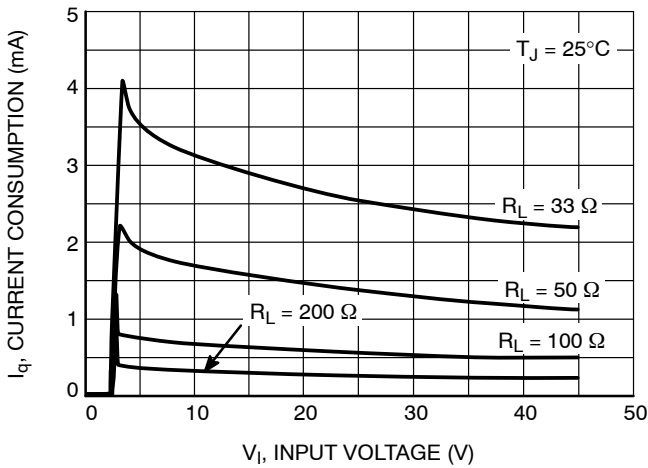


Figure 21. Current Consumption vs. Input Voltage

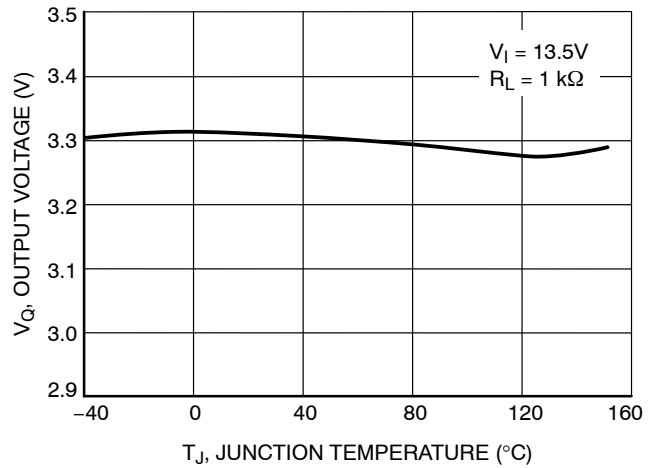


Figure 22. Output Voltage vs. Junction Temperature

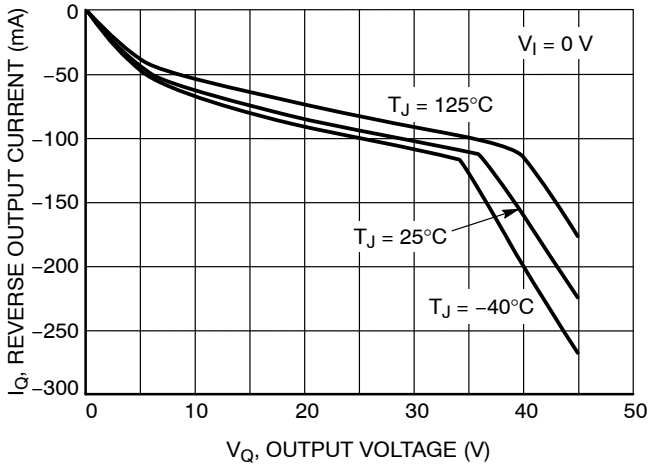


Figure 23. Reverse Output Current vs. Output Voltage

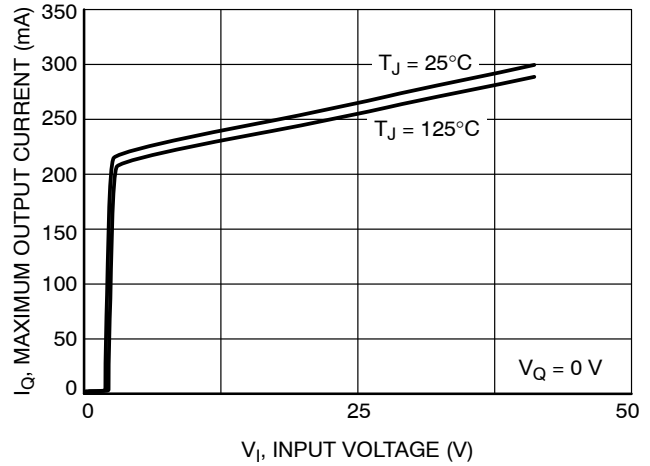


Figure 24. Maximum Output Current vs. Input Voltage

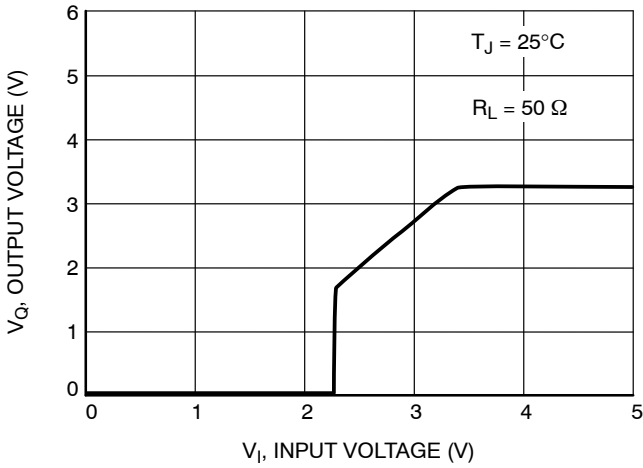


Figure 25. Output Voltage at Input Voltage Extremes

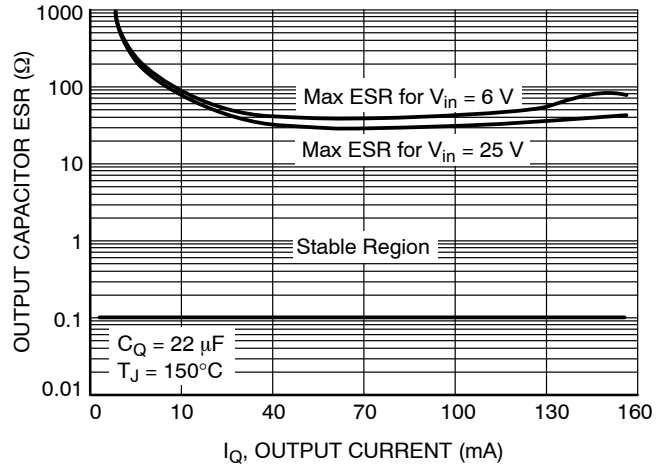


Figure 26. 3.3 V Output Stability with Output Capacitor ESR

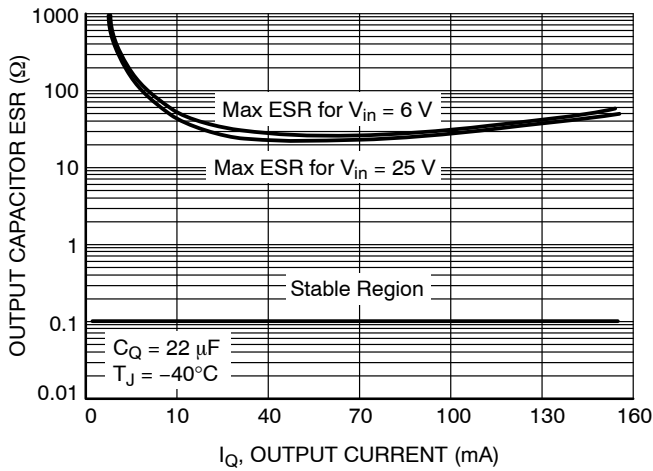


Figure 27. 3.3 V Output Stability with Output Capacitor ESR

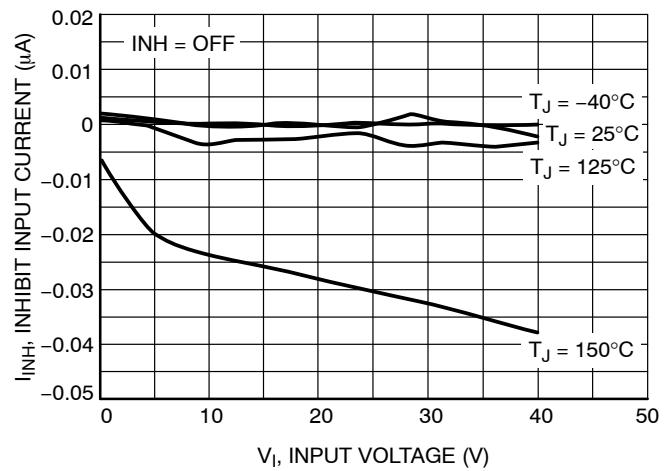


Figure 28. Inhibit Input Current at Input Voltage Extremes

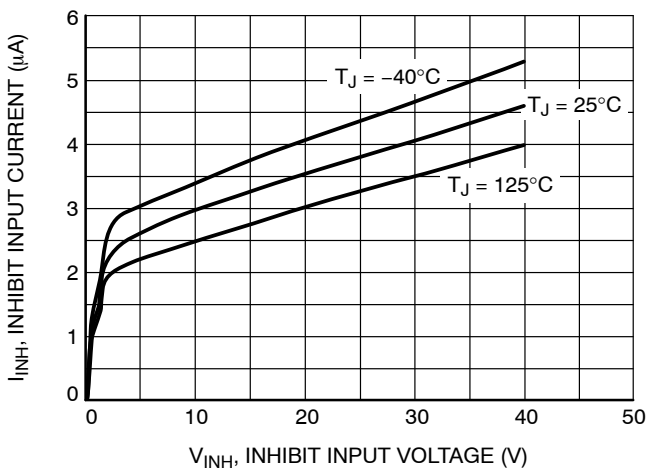


Figure 29. Inhibit Input Current at Inhibit Input Voltage Extremes

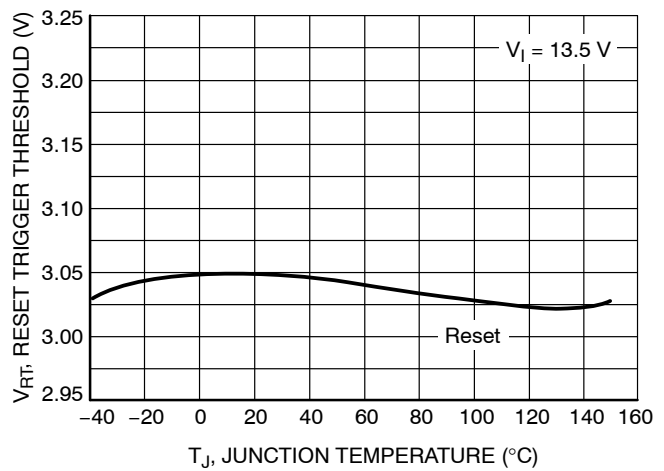


Figure 30. Reset Trigger Threshold vs. Junction Temperature

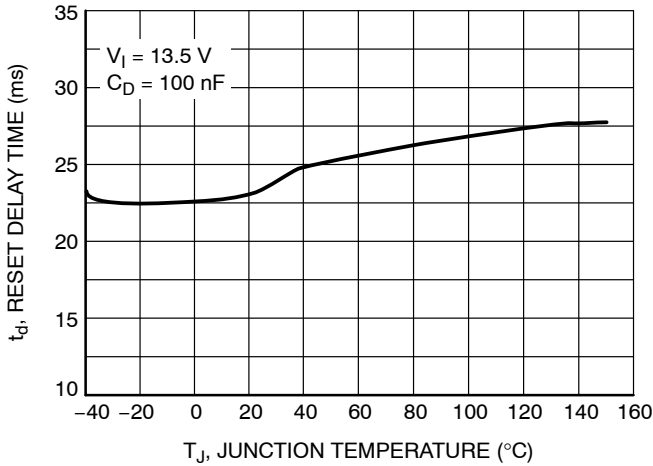


Figure 31. Reset Delay Time vs. Junction Temperature

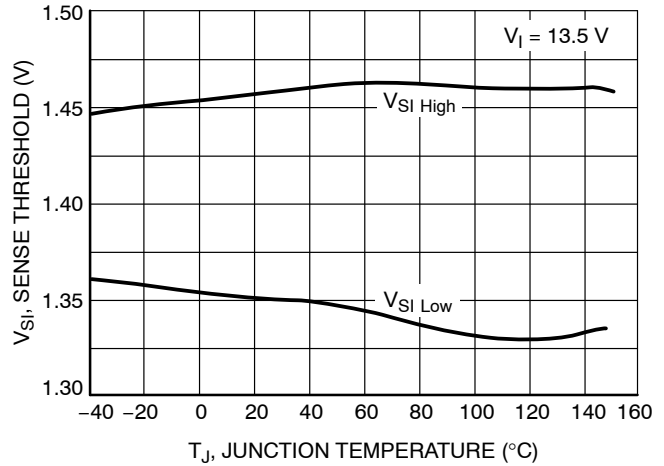


Figure 32. Sense Threshold vs. Junction Temperature

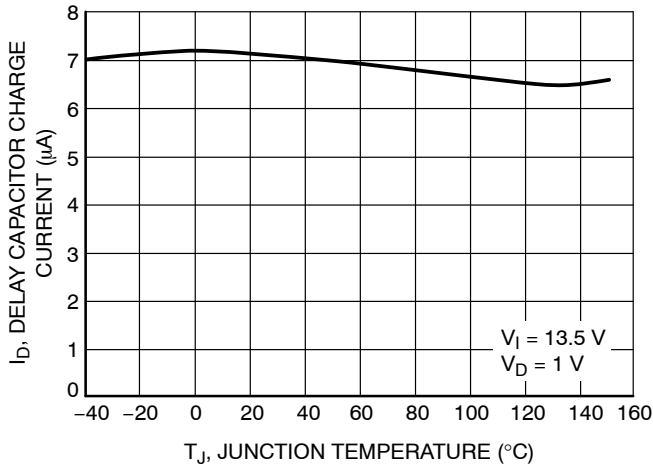


Figure 33. Delay Capacitor Charge Current vs. Junction Temperature

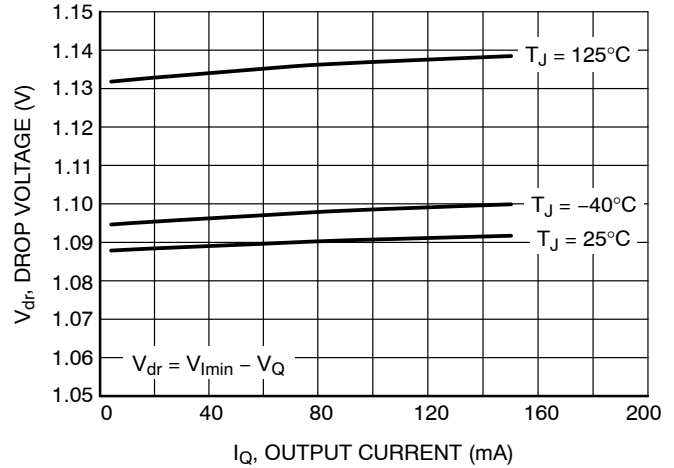


Figure 34. Drop Voltage vs. Output Current

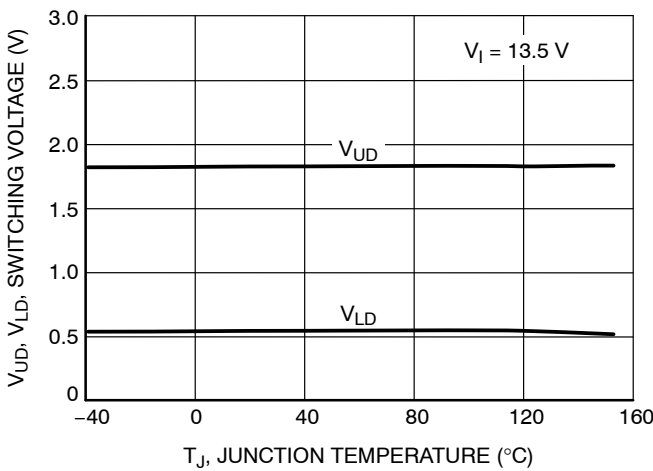


Figure 35. Switching Voltage  $V_{UD}$  and  $V_{LD}$  vs. Junction Temperature

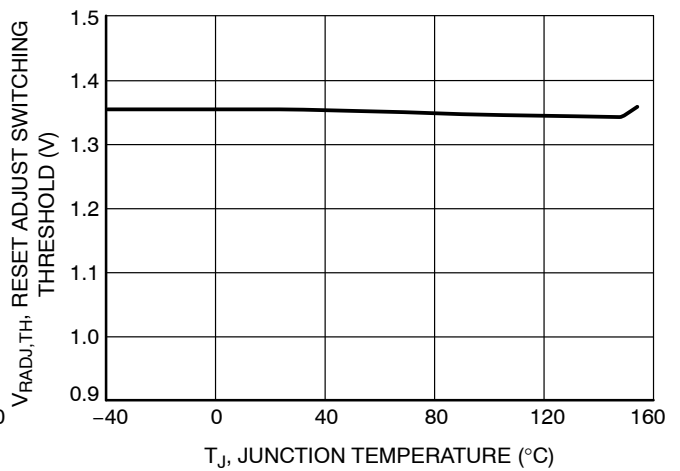


Figure 36. Reset Adjust Switching Threshold vs. Junction Temperature

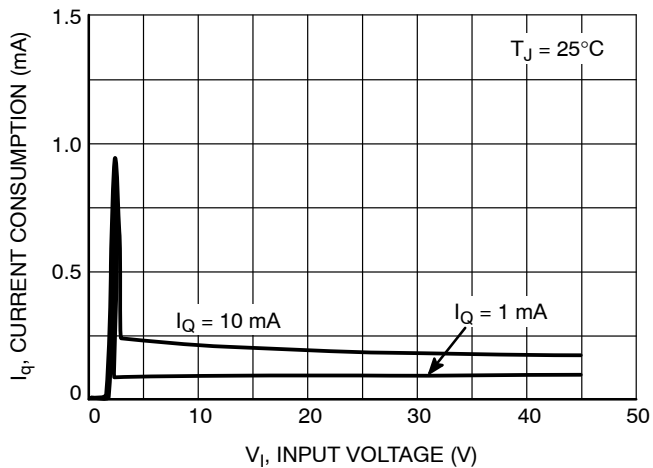


Figure 37. Current Consumption vs. Input Voltage

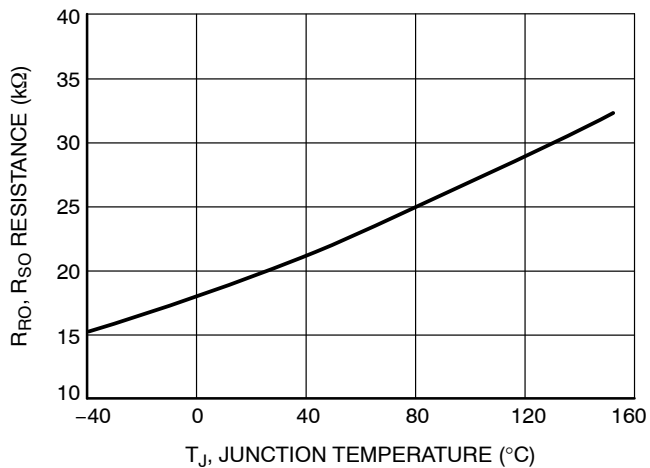


Figure 38.  $R_{RO}, R_{SO}$  Resistance vs. Junction Temperature

## APPLICATION DESCRIPTION

**NCV4299**

The NCV4299 is a family of precision micropower voltage regulators with an output current capability of 150 mA at 5.0 V and 3.3 V.

The output voltage is accurate within  $\pm 2\%$  with a maximum dropout voltage of 0.5 V at 100 mA. Low quiescent current is a feature drawing only 90  $\mu\text{A}$  with a 100  $\mu\text{A}$  load. This part is ideal for any and all battery operated microprocessor equipment.

Microprocessor control logic includes an active reset output RO (with delay), and a SI/SO monitor which can be used to provide an early warning signal to the microprocessor of a potential impending reset signal. The use of the SI/SO monitor allows the microprocessor to finish any signal processing before the reset shuts the microprocessor down. Internal output resistors on the RO and SO pins pulling up to the output pin Q reduce external component count. An inhibit function is available on the 14-lead part. With inhibit active, the regulator turns off and the device consumes less than 1.0  $\mu\text{A}$  of quiescent current.

The active reset circuit operates correctly at an output voltage as low as 1.0 V. The reset function is activated during the powerup sequence or during normal operation if the output voltage drops outside the regulation limits.

The reset threshold voltage can be decreased by the connection of an external resistor divider to the RADJ lead. The regulator is protected against reverse battery, short circuit, and thermal overload conditions. The device can withstand load dump transients making it suitable for use in automotive environments.

**NCV4299 Circuit Description**

The low dropout regulator in the NCV4299 uses a PNP pass transistor to give the lowest possible dropout voltage capability. The current is internally monitored to prevent oversaturation of the device and to limit current during over current conditions. Additional circuitry is provided to protect the device during overtemperature operation.

The regulator provides an output regulated to 2%.

Other features of the regulator include an undervoltage reset function and a sense circuit. The reset function has an adjustable time delay and an adjustable threshold level. The sense circuit trip level is adjustable and can be used as an early warning signal to the controller. An inhibit function that turns off the regulator and reduces the current consumption to less than 1.0  $\mu\text{A}$  is a feature available in the 14 pin package.

**Output Regulator**

The output is controlled by a precision trimmed reference. The PNP output has saturation control for regulation while the input voltage is low, preventing oversaturation. Current limit and voltage monitors complement the regulator design to give safe operating signals to the processor and control circuits.

**Stability Considerations**

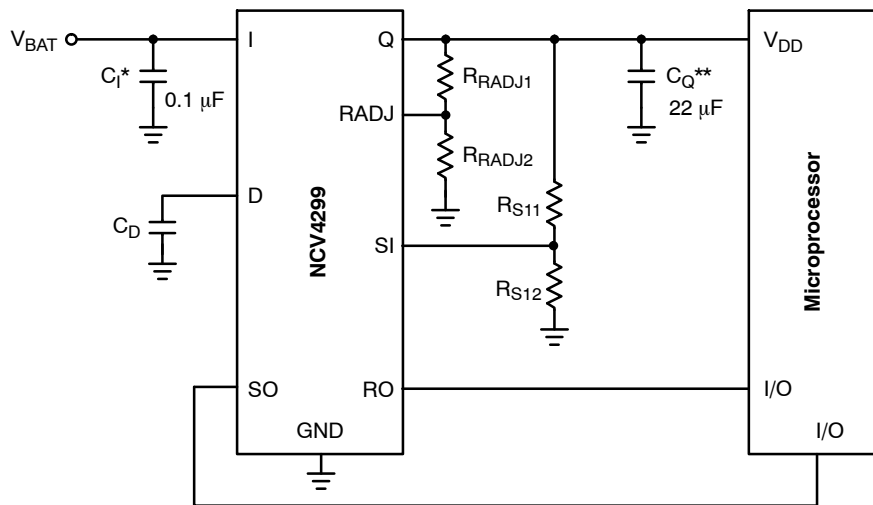
The input capacitor  $C_1$  is necessary for compensating input line reactance. Possible oscillations caused by input inductance and input capacitance can be damped by using a resistor of approximately 1.0  $\Omega$  in series with  $C_1$ .

The output or compensation capacitor helps determine three main characteristics of a linear regulator: startup delay, load transient response and loop stability.

The capacitor value and type should be based on cost, availability, size and temperature constraints. A tantalum or aluminum electrolytic capacitor is best, since a film or ceramic capacitor with almost zero ESR can cause instability. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures ( $-25^\circ\text{C}$  to  $-40^\circ\text{C}$ ), both the value and ESR of the capacitor will vary considerably. The capacitor manufacturer's data sheet usually provides this information.

The value for the output capacitor  $C_O$  shown in Figures 39 and 40 should work for most applications, however, it is not necessarily the optimized solution. Stability is guaranteed at values  $C_O \geq 22 \mu\text{F}$  and an  $\text{ESR} \leq 5.0 \Omega$  within the operating temperature range. Actual limits are shown in a graph in the typical performance characteristics section.

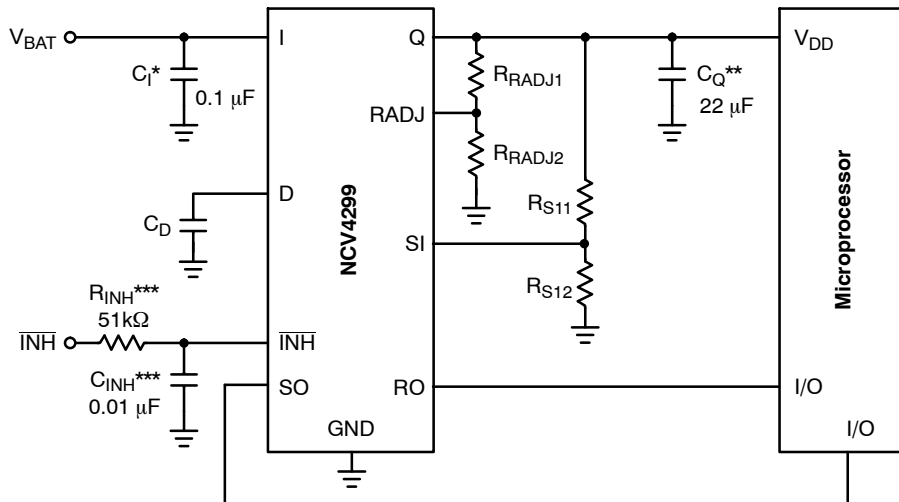
## NCV4299



\* $C_I$  required if regulator is located far from the power supply filter.

\*\* $C_Q$  required for stability. Cap must operate at minimum temperature expected.

**Figure 39. Test and Application Circuit Showing all Compensation and Sense Elements for the 8 Pin Package Part**



\* $C_I$  required if regulator is located far from the power supply filter.

\*\* $C_Q$  required for stability. Cap must operate at minimum temperature expected.

\*\*\*This RC filter is only required when transients with slew rate in excess of 10 V/ms may be present on the  $\overline{\text{INH}}$  voltage source during operation. The filter is not required when  $\overline{\text{INH}}$  is connected to a noise-free DC voltage.

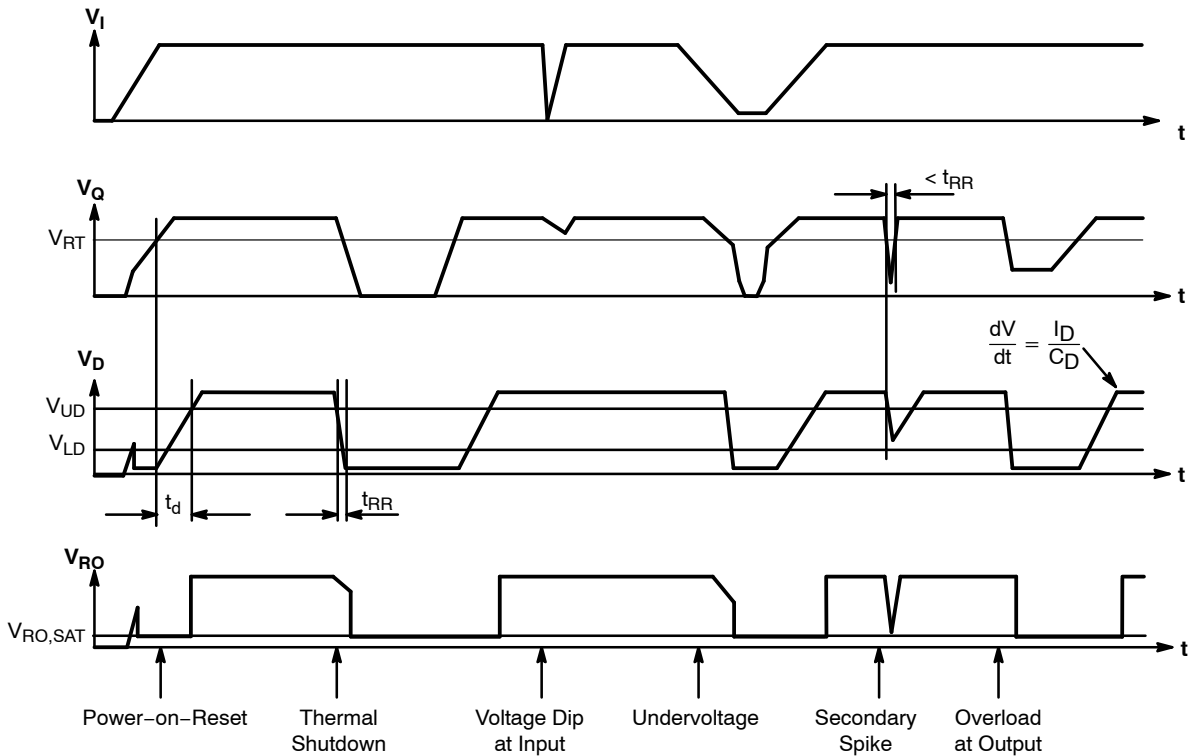
**Figure 40. Test and Application Circuit Showing all Compensation and Sense Elements for the 14 Pin Package Part with Inhibit Function**

**Reset Output (RO)**

A reset signal, Reset Output (RO, low voltage) is generated as the IC powers up. After the output voltage  $V_Q$  increases above the reset threshold voltage  $V_{RT}$ , the delay timer D is started. When the voltage on the delay timer  $V_D$  passes  $V_{UD}$ , the reset signal RO goes high. A discharge of the delay timer ( $V_D$ ) is started when  $V_Q$  drops and stays below the reset threshold voltage  $V_{RT}$ . When the voltage of

the delay timer ( $V_D$ ) drops below the lower threshold voltage  $V_{LD}$ , the reset output voltage  $V_{RO}$  is brought low to reset the processor.

The reset output RO is an open collector NPN transistor, controlled by a low voltage detection circuit. The circuit is functionally independent of the rest of the IC, thereby guaranteeing that RO is valid for  $V_Q$  as low as 1.0 V.



**Figure 41. Reset Timing Diagram**

**Reset Adjust (RADJ)**

The reset threshold  $V_{RT}$  can be decreased from a typical value of 4.64 V to as low as 3.5 V by using an external voltage divider connected from the Q lead to the pin RADJ, as shown in Figures 39 and 40. The resistor divider keeps the voltage above the  $V_{RADJ,TH}$  (typ. 1.36 V), for the desired input voltages and overrides the internal threshold detector. Adjust the voltage divider according to the following relationship:

$$V_{THRES} = V_{RADJ,TH} \cdot (R_{ADJ1} + R_{ADJ2}) / R_{ADJ2} \quad (\text{eq. 1})$$

If the reset adjust option is not needed, the RADJ-pin should be connected to GND causing the reset threshold to go to its default value (typ. 4.64 V).

**Reset Delay (D)**

The reset delay circuit provides a delay (programmable by capacitor  $C_D$ ) on the reset output RO lead. The delay lead D provides charge current  $I_D$  (typically 7.1  $\mu\text{A}$ ) to the external delay capacitor  $C_D$  during the following times:

1. During Powerup (once the regulation threshold has been exceeded).
2. After a reset event has occurred and the device is back in regulation. The delay capacitor is set to discharge when the regulation ( $V_{RT}$ , reset threshold voltage) has been violated. When the delay capacitor discharges to down to  $V_{LD}$ , the reset signal RO pulls low.

**Setting the Delay Time**

The delay time is set by the delay capacitor  $C_D$  and the charge current  $I_D$ . The time is measured by the delay capacitor voltage charging from the low level of  $V_{D,sat}$  to the higher level  $V_{UD}$ . The time delay follows the equation:

$$t_d = [C_D (V_{UD} - V_{D,sat})] / I_D \quad (\text{eq. 2})$$

Example:

Using  $C_D = 100 \text{ nF}$ .

Use the typical value for  $V_{D,sat} = 0.1 \text{ V}$ .

Use the typical value for  $V_{UD} = 1.85 \text{ V}$ .

Use the typical value for Delay Charge Current  $I_D = 7.1 \text{ }\mu\text{A}$ .

$$t_d = [100 \text{ nF}(1.85 - 0.1 \text{ V})] / 7.1 \text{ }\mu\text{A} = 24.6 \text{ ms} \quad (\text{eq. 3})$$

When the output voltage  $V_Q$  drops below the reset threshold voltage  $V_{RT}$ , the voltage on the delay capacitor  $V_D$  starts to drop. The time it takes to drop below the lower threshold voltage of  $V_{LD}$  is the reset reaction time,  $t_{RR}$ . This time is typically  $2.2 \text{ }\mu\text{s}$  for a delay capacitor of  $0.1 \text{ }\mu\text{F}$ . The reset reaction time can be estimated from the following relationship:

$$t_{RR} = 22 \text{ ns/nF} \times C_D \quad (\text{eq. 4})$$

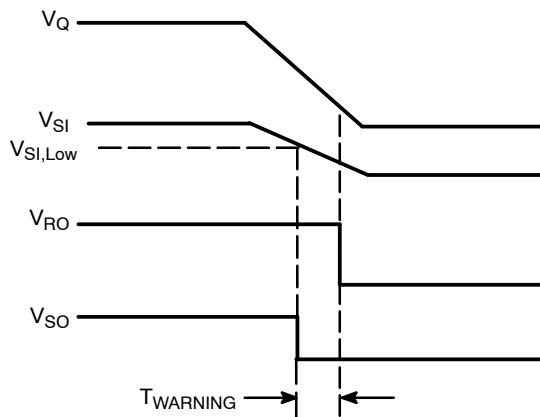


Figure 42. SO Warning Timing Waveform

**Calculating Power Dissipation in a Single Output Linear Regulator**

The maximum power dissipation for a single output regulator is:

$$P_{D(max)} = [V_{I(max)} - V_{Q(min)}] I_{Q(max)} + V_{I(max)} I_q \quad (\text{eq. 5})$$

where:

$V_{I(max)}$  is the maximum input voltage,

$V_{Q(min)}$  is the minimum output voltage,

$I_{Q(max)}$  is the maximum output current for the application,

and

**Sense Input (SI)/Sense Output (SO) Voltage Monitor**

An on-chip comparator is available to provide early warning to the microprocessor of a possible reset signal. The reset signal typically turns the microprocessor off instantaneously. This can cause unpredictable results with the microprocessor. The signal received from the SO pin will allow the microprocessor time to complete its present task before shutting down. This function is performed by a comparator referenced to the band gap voltage. The actual trip point can be programmed externally using a resistor divider to the input monitor (SI) (Figures 39 and 40). The typical threshold is  $1.35 \text{ V}$  on the SI Pin.

**Signal Output**

Figure 42 shows the SO Monitor waveforms as a result of the circuits depicted in Figures 39 and 40. As the output voltage  $V_Q$  falls, the monitor threshold  $V_{SI,Low}$  is crossed. This causes the voltage on the SO output to go low sending a warning signal to the microprocessor that a reset signal may occur in a short period of time.  $T_{WARNING}$  is the time the microprocessor has to complete the function it is currently working on and get ready for the reset shutdown signal.

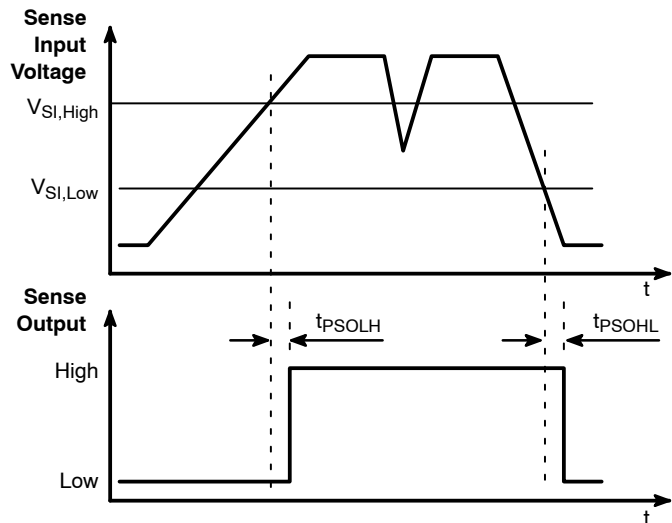


Figure 43. Sense Timing Diagram

$I_q$  is the quiescent current the regulator consumes at  $I_{Q(max)}$ .

Once the value of  $P_{D(max)}$  is known, the maximum permissible value of  $R_{\theta JA}$  can be calculated:

$$R_{\theta JA} = (150^\circ\text{C} - T_A) / P_D \quad (\text{eq. 6})$$

The value of  $R_{\theta JA}$  can then be compared with those in the package section of the data sheet. Those packages with  $R_{\theta JA}$ 's less than the calculated value in Equation 6 will keep the die temperature below  $150^\circ\text{C}$ . In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

## Heatsinks

A heatsink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of  $R_{\theta JA}$ :

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CS} + R_{\theta SA} \quad (\text{eq. 7})$$

where:

$R_{\theta JC}$  = the junction-to-case thermal resistance,  
 $R_{\theta CS}$  = the case-to-heatsink thermal resistance, and  
 $R_{\theta SA}$  = the heatsink-to-ambient thermal resistance.

$R_{\theta JC}$  appears in the package section of the data sheet. Like  $R_{\theta JA}$ , it too is a function of package type.  $R_{\theta CS}$  and  $R_{\theta SA}$  are functions of the package type, heatsink and the interface between them. These values appear in heatsink data sheets of heatsink manufacturers. Thermal, mounting, and heatsinking are discussed in the ON Semiconductor application note AN1040/D, available on the ON Semiconductor website.

SOIC 8 LEAD

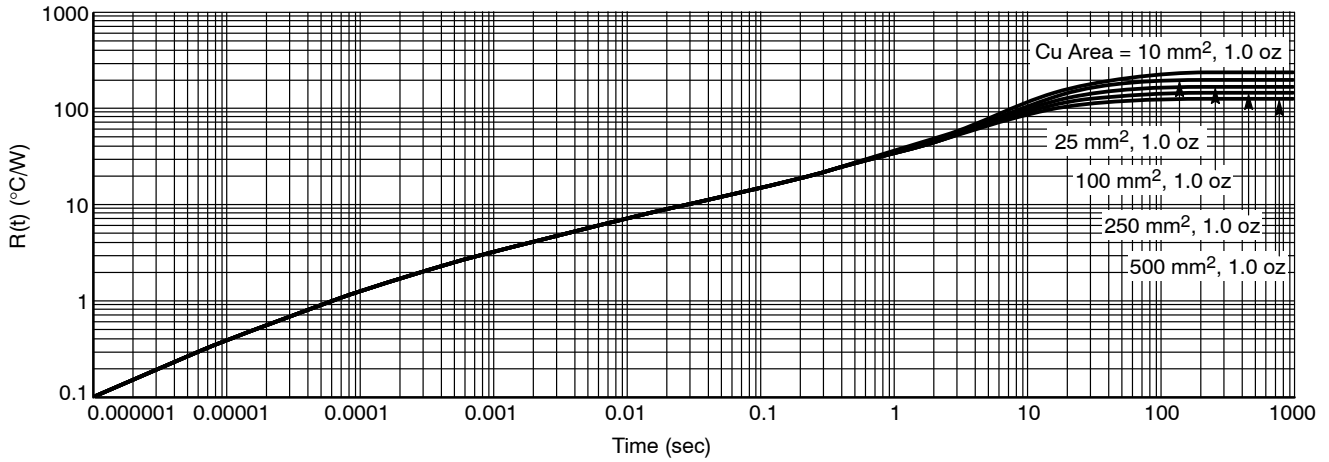


Figure 44. Transient Thermal Response Simulation to a Single Pulse 1 oz (Log-Log)

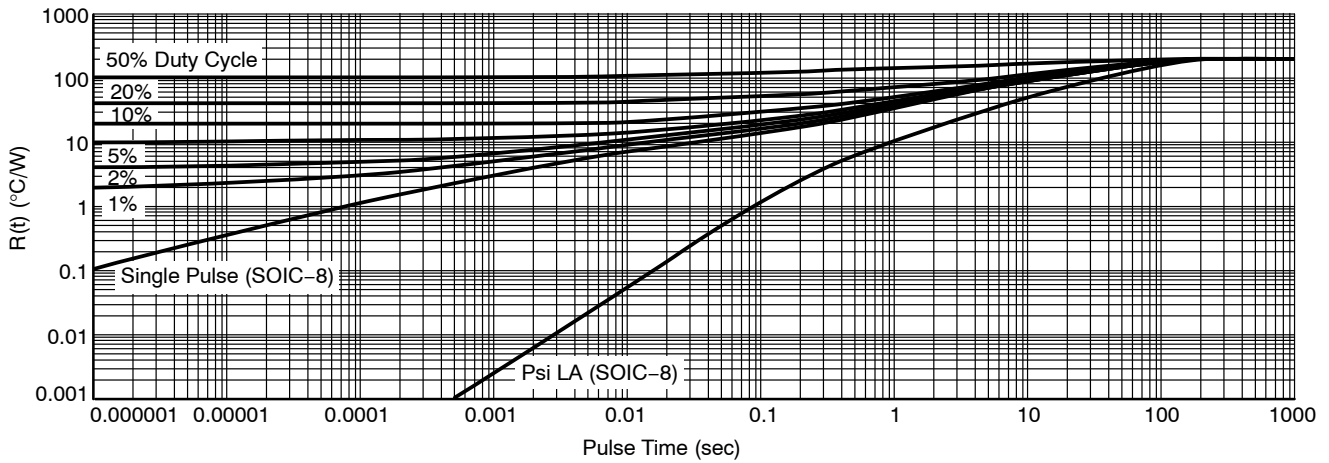


Figure 45. Transient Thermal Response Simulation to a Single Pulse with Duty Cycles Applied (Log-Log)  
(PCB = 50 mm<sup>2</sup> 1 oz)

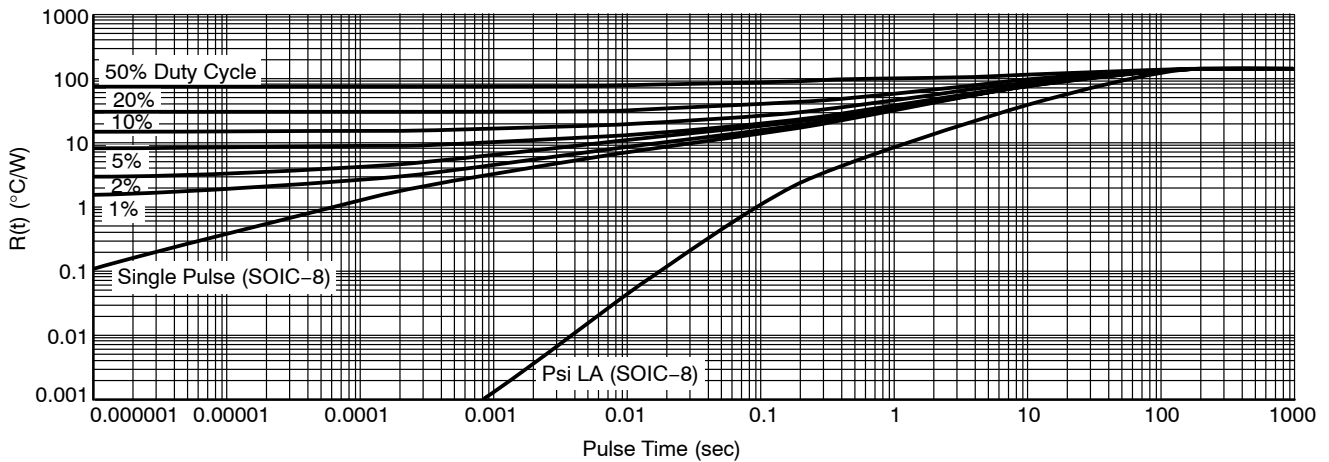


Figure 46. Transient Thermal Response Simulation to a Single Pulse with Duty Cycles Applied (Log-Log)  
(PCB = 250 mm<sup>2</sup> 1 oz)

SOIC 14 LEAD

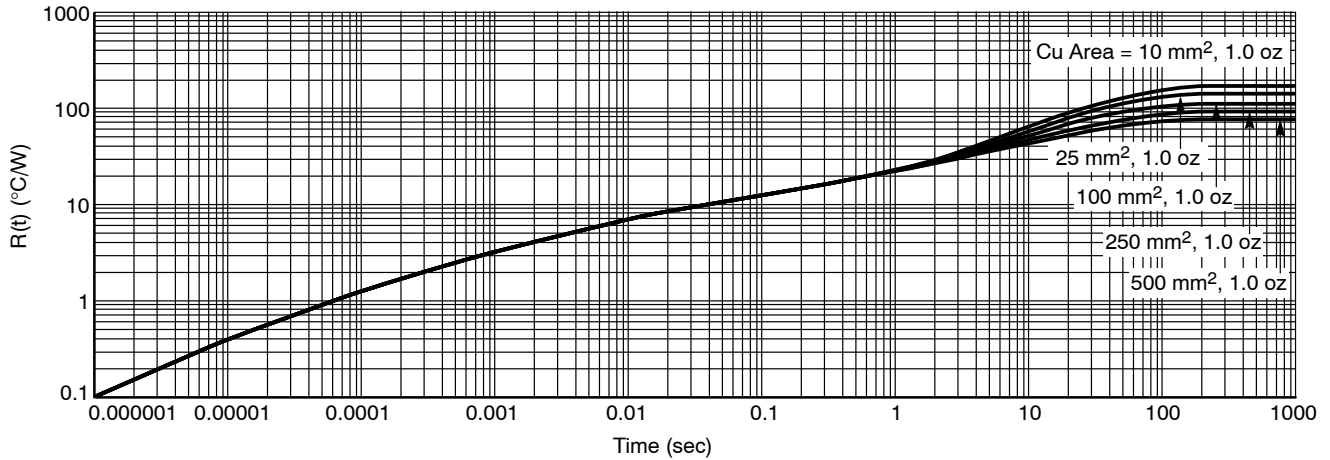


Figure 47. Transient Thermal Response Simulation to a Single Pulse 1 oz (Log-Log)

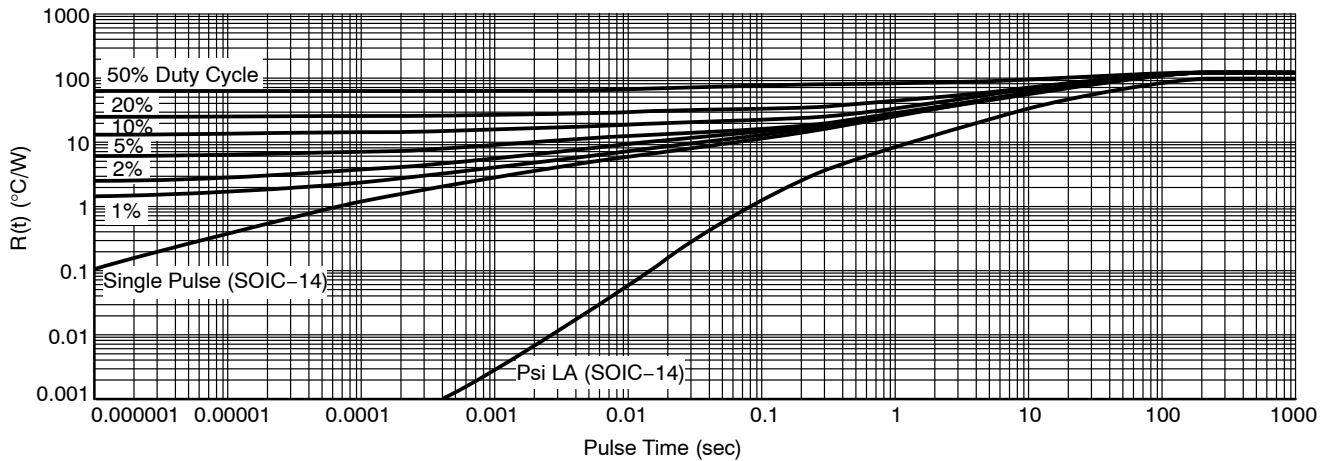


Figure 48. Transient Thermal Response Simulation to a Single Pulse with Duty Cycles Applied (Log-Log)  
(PCB = 50 mm<sup>2</sup> 1 oz)

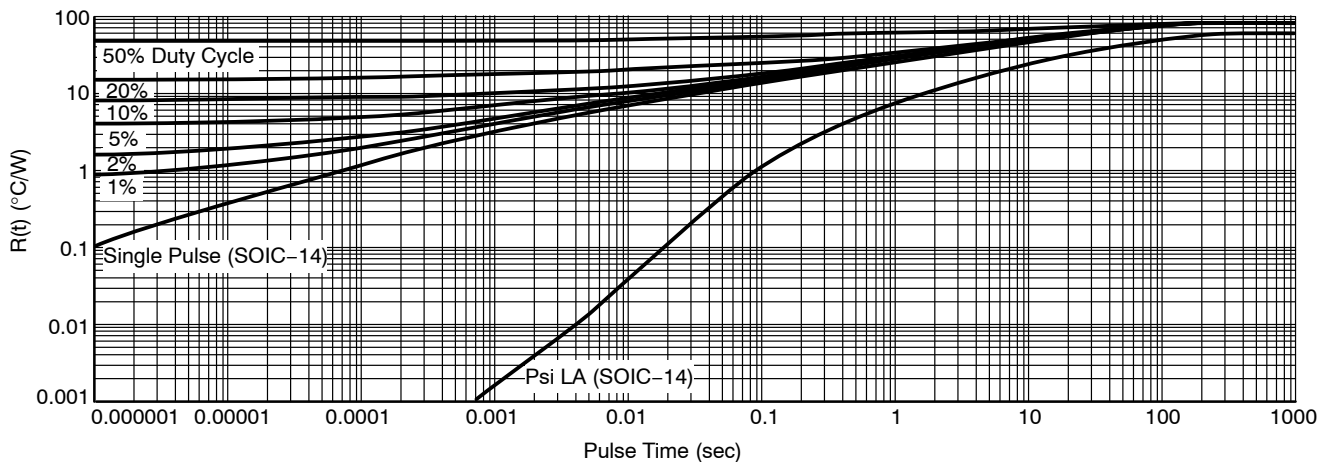


Figure 49. Transient Thermal Response Simulation to a Single Pulse with Duty Cycles Applied (Log-Log)  
(PCB = 250 mm<sup>2</sup> 1 oz)

# NCV4299

## ORDERING INFORMATION

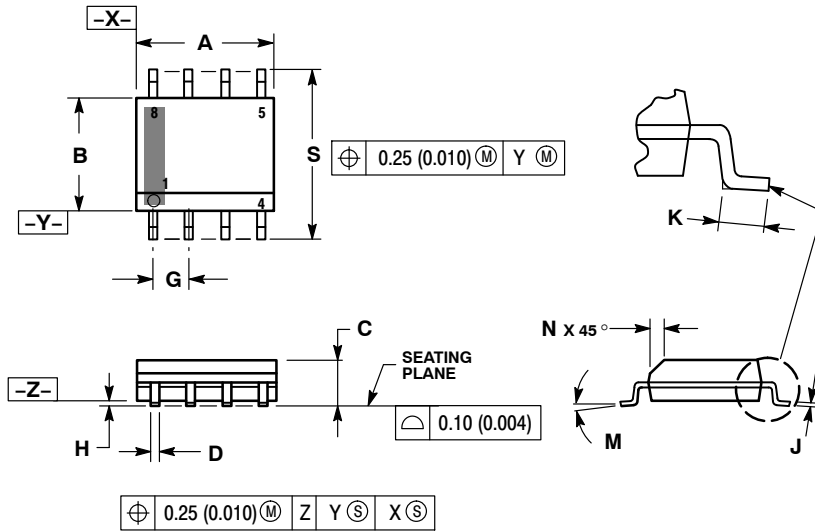
Device	Package	Shipping†
NCV4299D1G	SO-8 (Pb-Free)	98 Units/Rail
NCV4299D1R2G	SO-8 (Pb-Free)	2500 Tape & Reel
NCV4299D2G	SO-14 (Pb-Free)	55 Units/Rail
NCV4299D2R2G	SO-14 (Pb-Free)	2500 Tape & Reel
NCV4299D233G	SO-14 (Pb-Free)	55 Units/Rail
NCV4299D233R2G	SO-14 (Pb-Free)	2500 Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

# NCV4299

## PACKAGE DIMENSIONS

SOIC-8 NB  
CASE 751-07  
ISSUE AJ

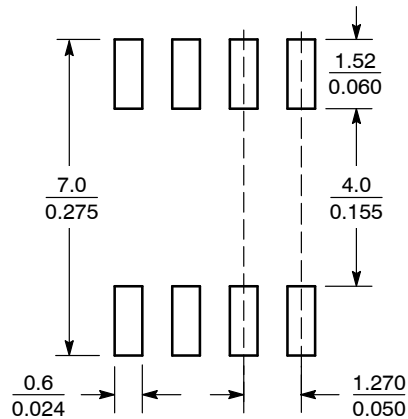


### NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	4.80	5.00	0.189	0.197
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.053	0.069
D	0.33	0.51	0.013	0.020
G	1.27 BSC		0.050 BSC	
H	0.10	0.25	0.004	0.010
J	0.19	0.25	0.007	0.010
K	0.40	1.27	0.016	0.050
M	0°	8°	0°	8°
N	0.25	0.50	0.010	0.020
S	5.80	6.20	0.228	0.244

### SOLDERING FOOTPRINT\*



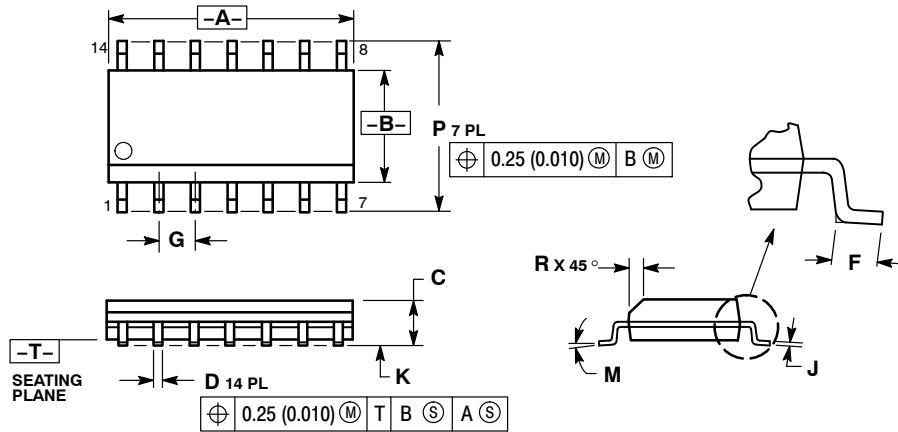
SCALE 6:1 ( $\frac{\text{mm}}{\text{inches}}$ )

\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

# NCV4299

## PACKAGE DIMENSIONS

SOIC-14  
CASE 751A-03  
ISSUE H

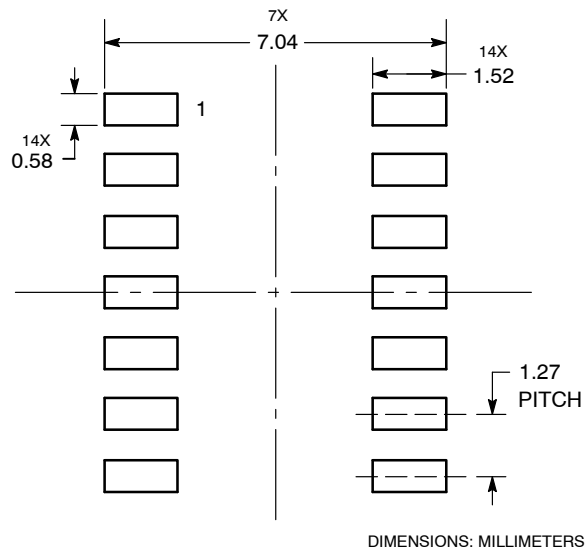


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2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	8.55	8.75	0.337	0.344
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050 BSC	
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	0°	7°	0°	7°
P	5.80	6.20	0.228	0.244
R	0.25	0.50	0.010	0.019

### SOLDERING FOOTPRINT



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