



**THE DATASHEET OF
AP7335-39SNG-7**



Description

The AP7335 is a 300mA, adjustable and fixed output voltage, low dropout linear regulator. The device included pass element, error amplifier, band-gap, current limit and thermal shutdown circuitry. The device is turned on when EN pin is set to logic high level.

The characteristics of low dropout voltage and low quiescent current make it suitable for low power applications, for example, battery powered devices. The typical quiescent current is approximately 35µA. Built-in current-limit and thermal-shutdown functions prevent IC from damage in fault conditions.

This device is available with adjustable output from 0.8V to 5.0V, and fixed version with 0.8V, 1.0V, 1.2V, 1.5V, 1.8V, 2.0V, 2.5V, 2.7V, 2.8V, 3.0V, 3.3V and 3.9V outputs. Please contact your local sales office for any other voltage options.

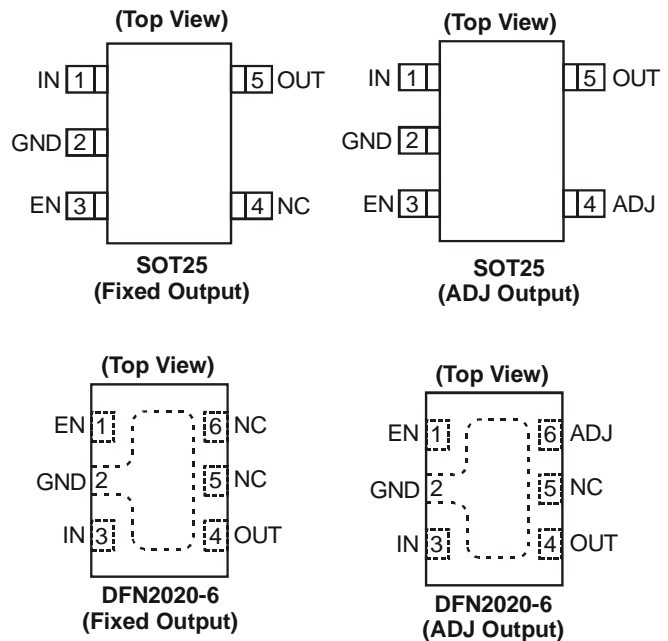
The AP7335 is available in SOT25 and DFN2020-6 packages.

Features

- 300mA Low Dropout Regulator with EN
- Very low I_Q : 35µA
- Wide input voltage range: 2V to 6V
- Wide adjustable output: 0.8V to 5.0V
- Fixed output options: 0.8V to 3.9V (0.1V step size possible)
- High PSRR: 65dB at 1kHz
- Fast start-up time: 220µs
- Stable with low ESR, 1µF ceramic output capacitor
- Excellent Load/Line Transient Response
- Low dropout: 150mV at 300mA
- Current limit and short circuit protection
- Thermal shutdown protection
- Ambient temperature range: -40°C to 85°C
- SOT25, and DFN2020-6: Available in "Green" Molding Compound (No Br, Sb)
- Lead Free Finish/ RoHS Compliant (Note 1)

Note: 1. EU Directive 2002/95/EC (RoHS). All applicable RoHS exemptions applied. Please visit our website at http://www.diodes.com/products/lead_free.html.

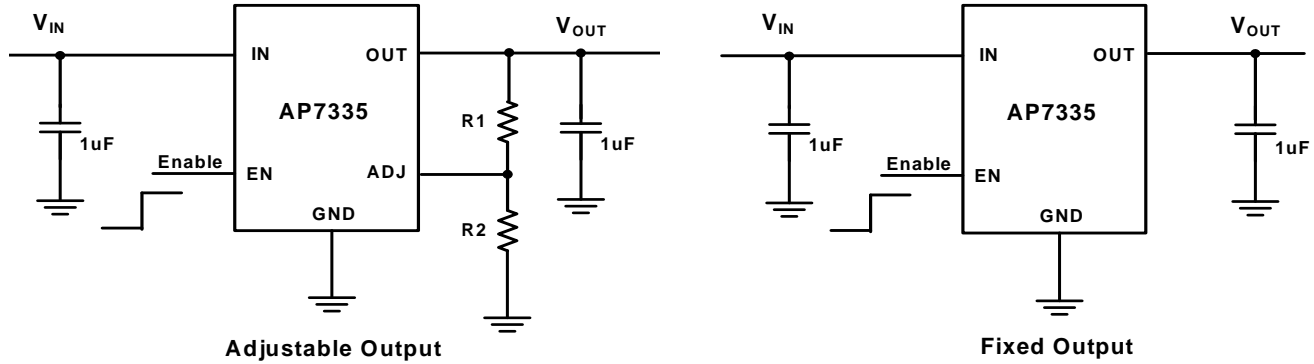
Pin Assignments



Applications

- Smart Phones
- MP3/MP4
- Battery-powered devices
- Bluetooth headset

Typical Application Circuit



$$V_{OUT} = V_{REF} \left(1 + \frac{R_1}{R_2} \right) \text{ where } R_2 \leq 80K\Omega$$

Pin Descriptions

Pin Name	Pin Number				Description
	SOT25 (fixed)	SOT25 (adj)	DFN2020-6 (fixed)	DFN2020-6 (adj)	
IN	1	1	3	3	Voltage input pin. Bypass to ground through at least 1µF MLCC capacitor
GND	2	2	2	2	Ground
EN	3	3	1	1	Enable input, active high
ADJ	-	4	-	6	Output feedback pin
NC	4	-	5, 6	5	No connection
OUT	5	5	4	4	Voltage output pin. Bypass to ground through 1µF MLCC capacitor

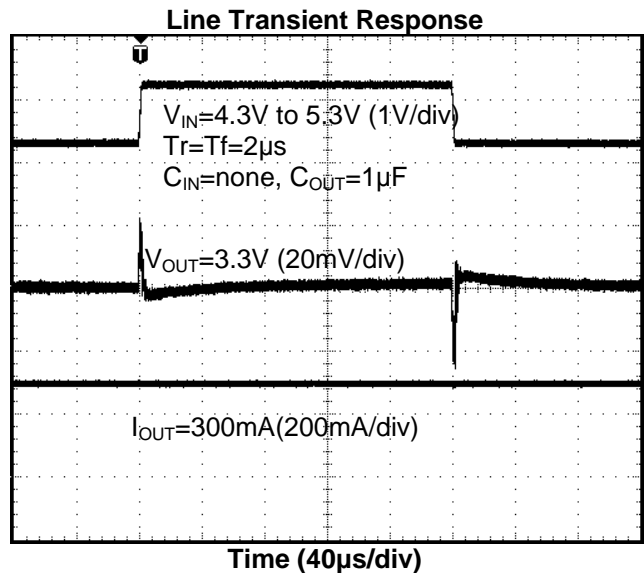
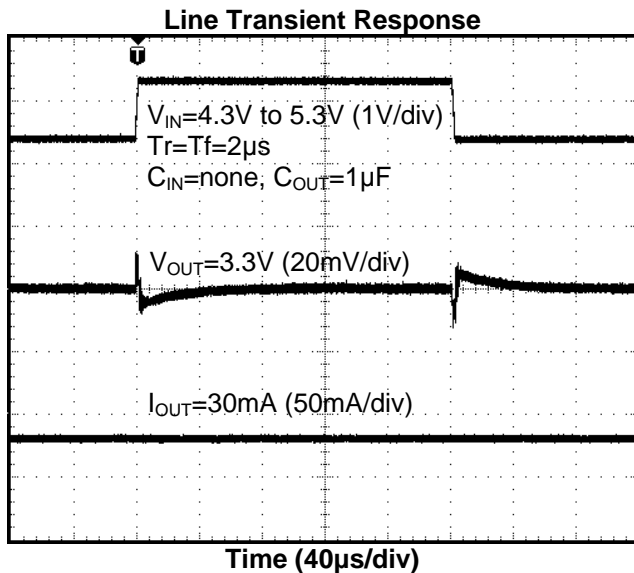
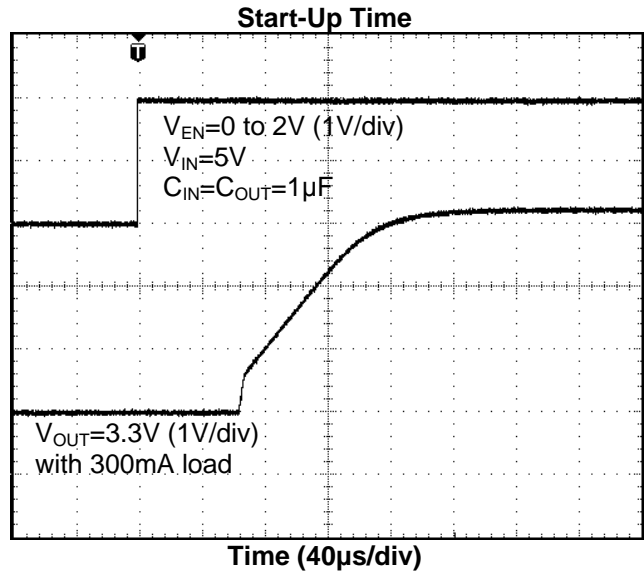
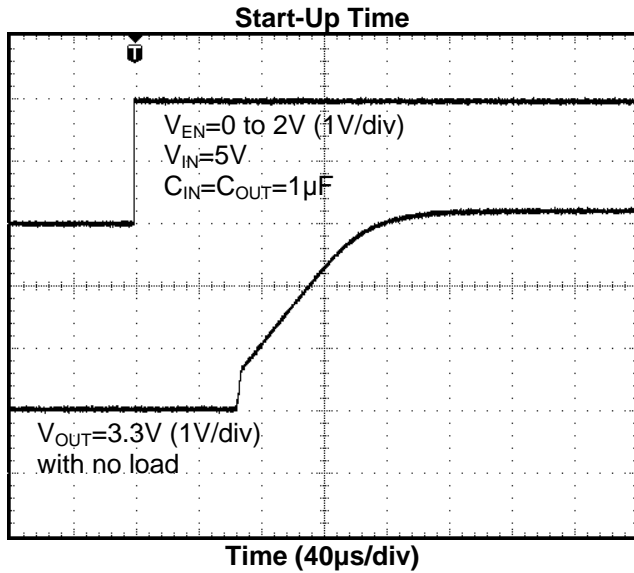
Electrical Characteristics

($T_A = 25^\circ\text{C}$, $V_{IN} = V_{OUT} + 1\text{V}$, $C_{IN} = 1\mu\text{F}$, $C_{OUT} = 1\mu\text{F}$, $V_{EN} = 2\text{V}$, unless otherwise stated)

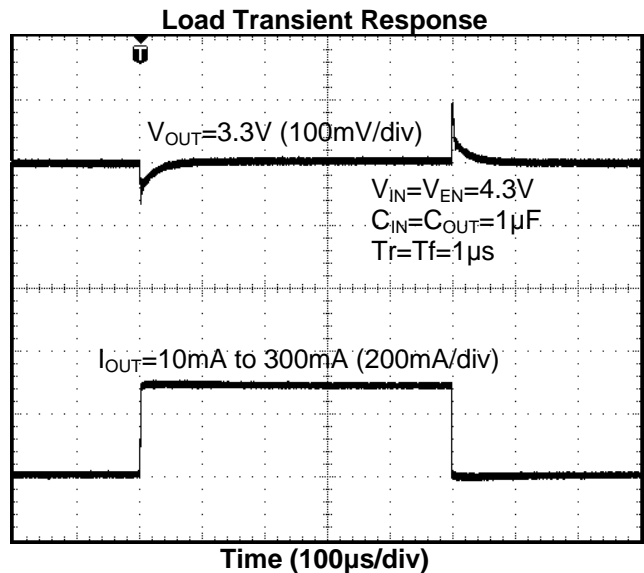
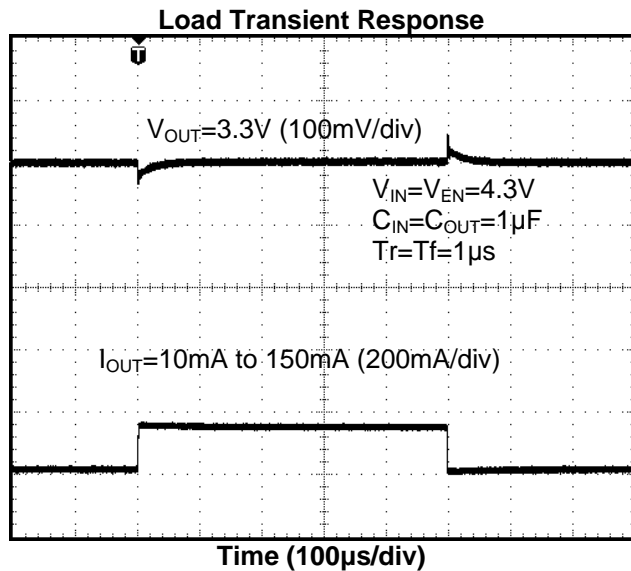
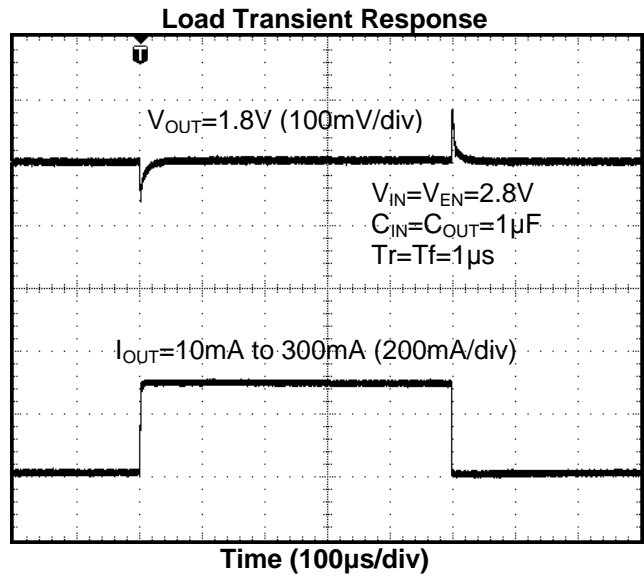
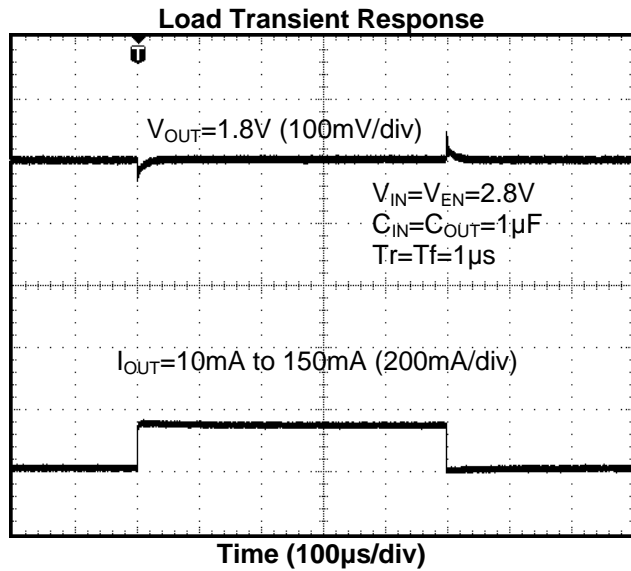
Symbol	Parameter	Test Conditions	Min	Typ.	Max	Unit
V_{REF}	ADJ Reference Voltage (Adjustable version)	$I_{OUT} = 0\text{mA}$		0.8		V
I_{ADJ}	ADJ Leakage (Adjustable version)			0.1	1	μA
V_{OUT}	Output Voltage Accuracy	$T_A = -40^\circ\text{C}$ to 85°C , $I_{OUT} = 10\%$ of $I_{OUT-Max}$	-2		2	%
$\frac{\Delta V_{OUT}}{\Delta V_{IN}}/V$	Line Regulation	$V_{IN} = (V_{OUT} + 1\text{V})$ to V_{IN-Max} , $V_{EN} = V_{IN}$, $I_{OUT} = 1\text{mA}$		0.02	0.20	%/V
$\frac{\Delta V_{OUT}}{V_{OUT}}$	Load Regulation	$V_{IN} = (V_{OUT} + 1\text{V})$ to V_{IN-Max} , $I_{OUT} = 1\text{mA}$ to 300mA	-0.6		0.6	%
$V_{Dropout}$	Dropout Voltage (Note 3)	$V_{OUT} < 2.5\text{V}$, $I_{OUT} = 300\text{mA}$		170	300	mV
		$V_{OUT} \geq 2.5\text{V}$, $I_{OUT} = 300\text{mA}$		150	200	
I_Q	Input Quiescent Current	$V_{EN} = V_{IN}$, $I_{OUT} = 0\text{mA}$		35	80	μA
I_{SHDN}	Input Shutdown Current	$V_{EN} = 0\text{V}$, $I_{OUT} = 0\text{mA}$		0.1	1	μA
I_{LEAK}	Input Leakage Current	$V_{EN} = 0\text{V}$, OUT grounded		0.1	1	μA
t_{ST}	Start-up Time	$V_{EN} = 0\text{V}$ to 2.0V in $1\mu\text{s}$, $I_{OUT} = 300\text{mA}$		220		μs
PSRR	PSRR (Note 4)	$V_{IN} = [V_{OUT} + 1\text{V}]V_{DC} + 0.5V_{ppAC}$, $f = 1\text{kHz}$, $I_{OUT} = 50\text{mA}$		65		dB
I_{SHORT}	Short-circuit Current	$V_{IN} = V_{IN-Min}$ to V_{IN-Max} , $V_{OUT} < 0.2\text{V}$ (fixed) or 25% of V_{OUT} (ADJ version)		140		mA
I_{LIMIT}	Current limit	$V_{IN} = V_{IN-Min}$ to V_{IN-Max} , $V_{OUT}/R_{OUT} = 1.2\text{A}$	400	600		mA
V_{IL}	EN Input Logic Low Voltage	$V_{IN} = V_{IN-Min}$ to V_{IN-Max}			0.4	V
V_{IH}	EN Input Logic High Voltage	$V_{IN} = V_{IN-Min}$ to V_{IN-Max}	1.4			V
I_{EN}	EN Input Current	$V_{IN} = 0\text{V}$ or V_{IN-Max}	-1		1	μA
T_{SHDN}	Thermal shutdown threshold			145		$^\circ\text{C}$
T_{HYS}	Thermal shutdown hysteresis			15		$^\circ\text{C}$
θ_{JA}	Thermal Resistance Junction-to-Ambient	SOT25 (Note 5)		187		$^\circ\text{C}/\text{W}$
		DFN2020-6 (Note 5)		251		

- Notes:
- Dropout voltage is the voltage difference between the input and the output at which the output voltage drops 2% below its nominal value. This parameter only applies to input voltages above minimum $V_{IN} = 2.0\text{V}$.
 - At $V_{IN} < 2.3\text{V}$, the PSRR performance may be reduced.
 - Test condition for all packages: Device mounted on FR-4 substrate PC board, 1oz copper, with minimum recommended pad layout.

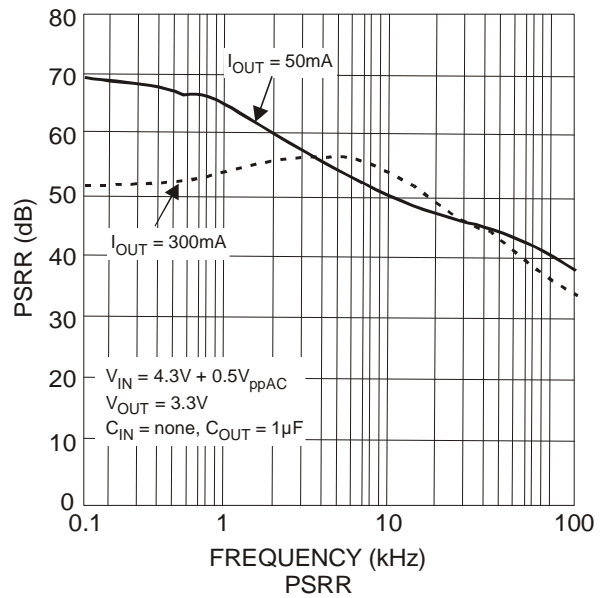
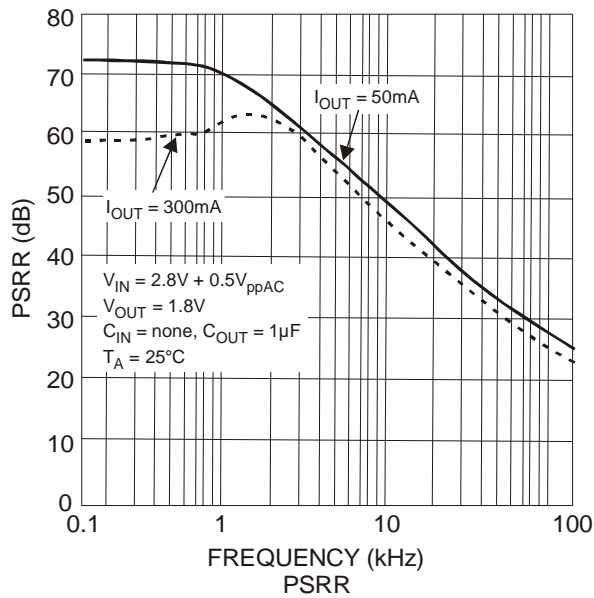
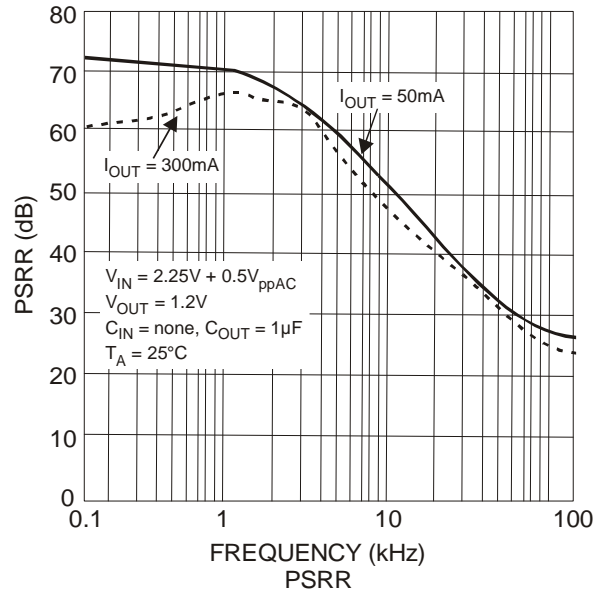
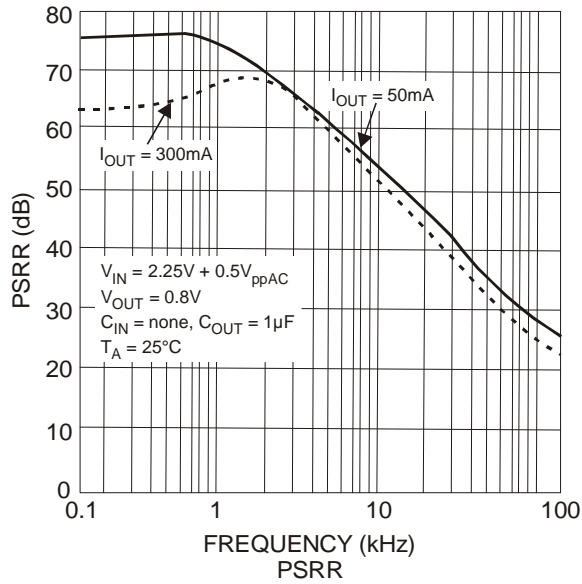
Typical Performance Characteristics



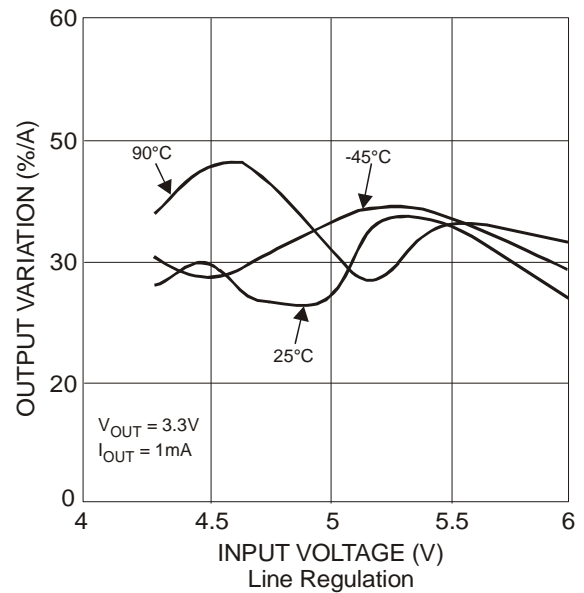
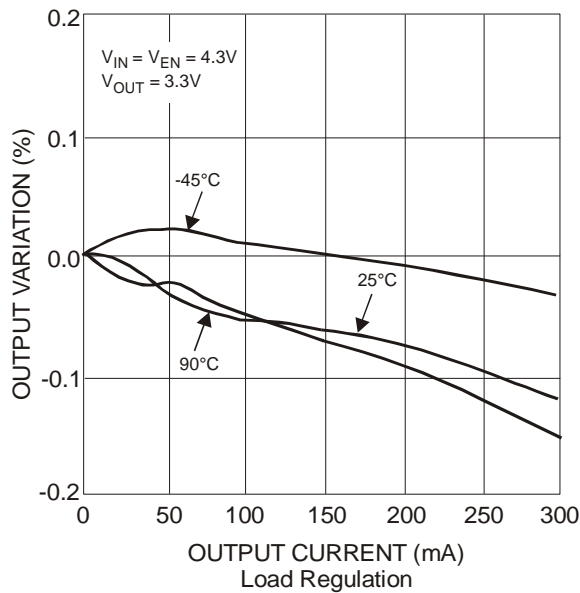
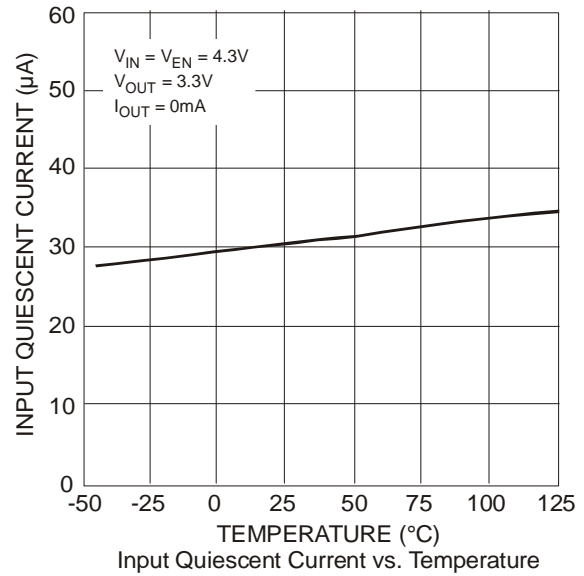
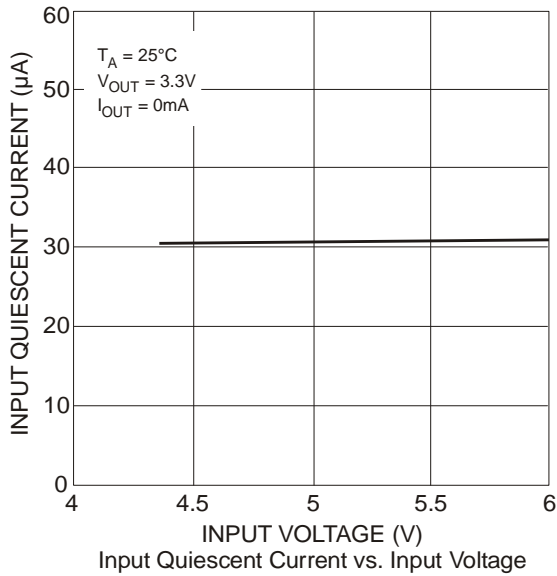
Typical Performance Characteristics (cont.)



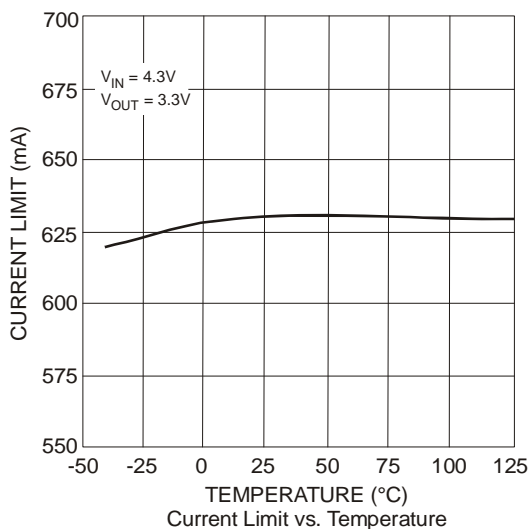
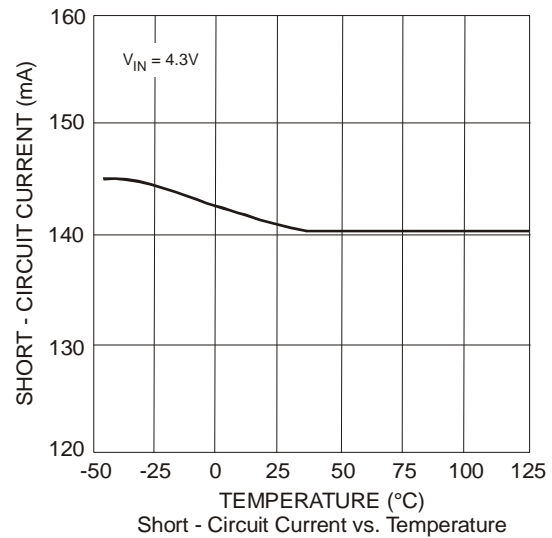
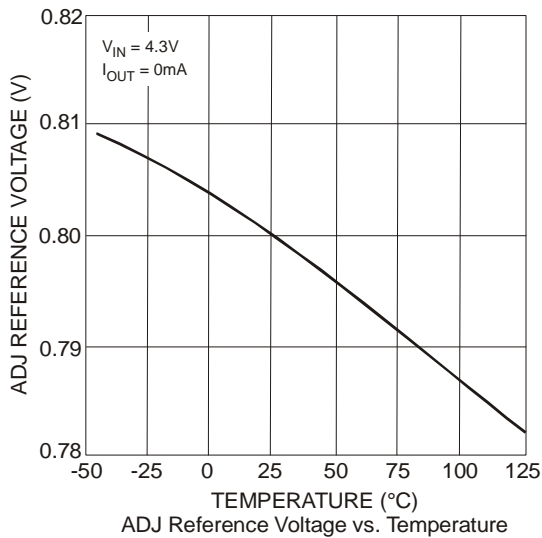
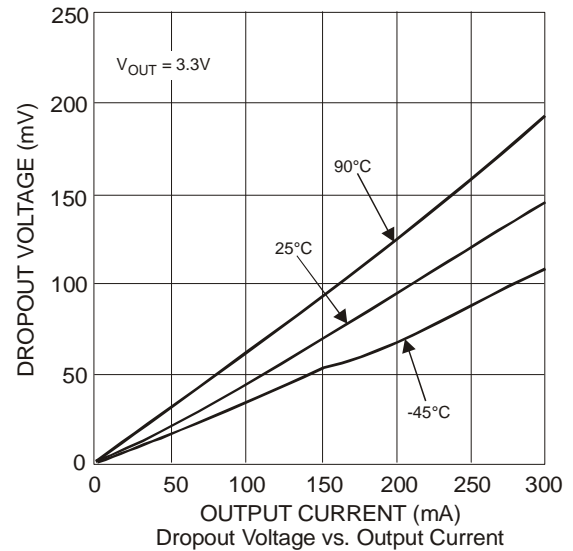
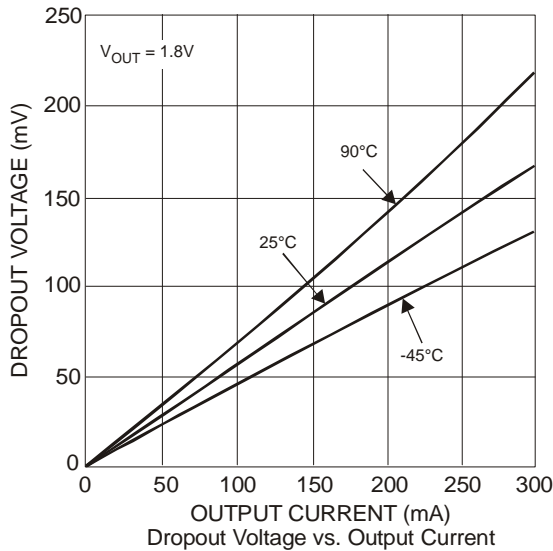
Typical Performance Characteristics (cont.)



Typical Performance Characteristics (cont.)



Typical Performance Characteristics (cont.)



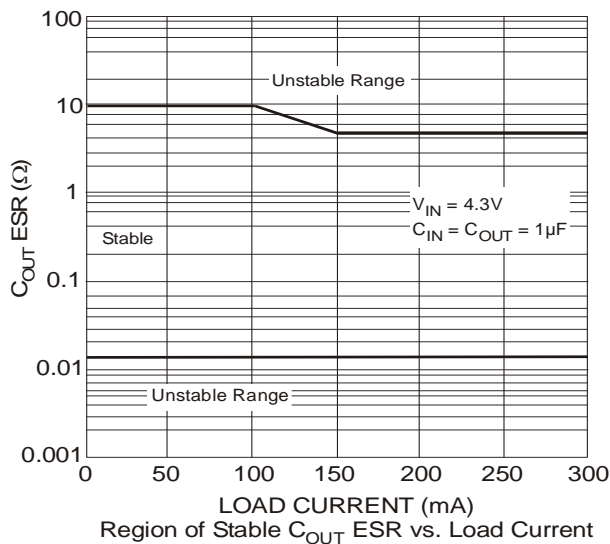
Application Notes

Input Capacitor

A 1µF ceramic capacitor is recommended between IN and GND pins to decouple input power supply glitch and noise. The amount of the capacitance may be increased without limit. This input capacitor must be located as close as possible to the device to assure input stability and reduce noise. For PCB layout, a wide copper trace is required for both IN and GND pins. A lower ESR capacitor type allows the use of less capacitance, while higher ESR type requires more capacitance.

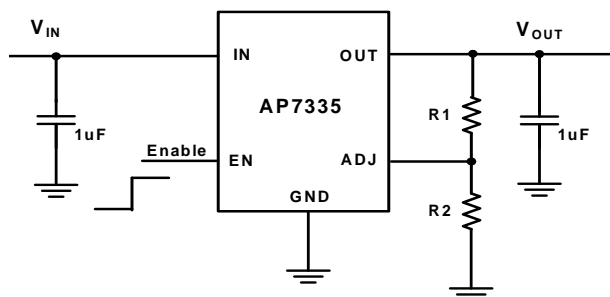
Output Capacitor

The output capacitor is required to stabilize and improve the transient response of the LDO. The AP7335 is stable with very small ceramic output capacitors. Using a ceramic capacitor value that is at least 1µF with ESR > 15mΩ on the output ensures stability. Higher capacitance values help to improve line and load transient response. The output capacitance may be increased to keep low undershoot and overshoot. Output capacitor must be placed as close as possible to OUT and GND pins.



Adjustable Operation

The AP7335 provides output voltage from 0.8V to 5.0V through external resistor divider as shown below.



The output voltage is calculated by:

$$V_{OUT} = V_{REF} \left(1 + \frac{R_1}{R_2} \right)$$

Where $V_{REF}=0.8V$ (the internal reference voltage)

Rearranging the equation will give the following that is used for adjusting the output to a particular voltage:

$$R_1 = R_2 \left(\frac{V_{OUT}}{V_{REF}} - 1 \right)$$

To maintain the stability of the internal reference voltage, R_2 need to be kept smaller than 80kΩ.

No Load Stability

Other than external resistor divider, no minimum load is required to keep the device stable. The device will remain stable and regulated in no load condition.

ON/OFF Input Operation

The AP7335 is turned on by setting the EN pin high, and is turned off by pulling it low. If this feature is not used, the EN pin should be tied to IN pin to keep the regulator output on at all time. To ensure proper operation, the signal source used to drive the EN pin must be able to swing above and below the specified turn-on/off voltage thresholds listed in the Electrical Characteristics section under V_{IL} and V_{IH} .

Current Limit Protection

When output current at OUT pin is higher than current limit threshold, the current limit protection will be triggered and clamp the output current to approximately 600mA to prevent over-current and to protect the regulator from damage due to overheating.

Short Circuit Protection

When OUT pin is short-circuit to GND, short circuit protection will be triggered and clamp the output current to approximately 140mA. This feature protects the regulator from over-current and damage due to overheating.

Thermal Shutdown Protection

Thermal protection disables the output when the junction temperature rises to approximately +145°C, allowing the device to cool down. When the junction temperature reduces to approximately +130°C the output circuitry is enabled again. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits the heat dissipation of the regulator, protecting it from damage due to overheating.

Application Notes

Ultra Fast Start-up

After enabled, the AP7335 is able to provide full power in as little as tens of microseconds, typically 220µs, without sacrificing low ground current. This feature will help load circuitry move in and out of standby mode in real time, eventually extend battery life for mobile phones and other portable devices.

Fast Transient Response

Fast transient response LDO can extend battery life. TDMA-based cell phone protocols such as Global System for Mobile Communications (GSM) have a transmit/receive duty factor of only 12.5 percent, enabling power savings by putting much of the baseband circuitry into standby mode in between transmit cycles. In baseband circuits, the load often transitions virtually instantaneously from 100µA to 100mA. To meet this load requirement, the LDO must react very quickly without a large voltage drop or overshoot — a requirement that cannot be met with conventional, general-purpose LDO.

The AP7335's fast transient response from 0 to 300mA provides stable voltage supply for fast DSP and GSM chipset with fast changing load.

Low Quiescent Current

The AP7335, consuming only around 35µA for all input range, provides great power saving in portable and low power applications.

Wide Output Range

The AP7335, with a wide output range of 0.8V to 5.0V, provides a versatile LDO solution for many portable applications.

Power Dissipation

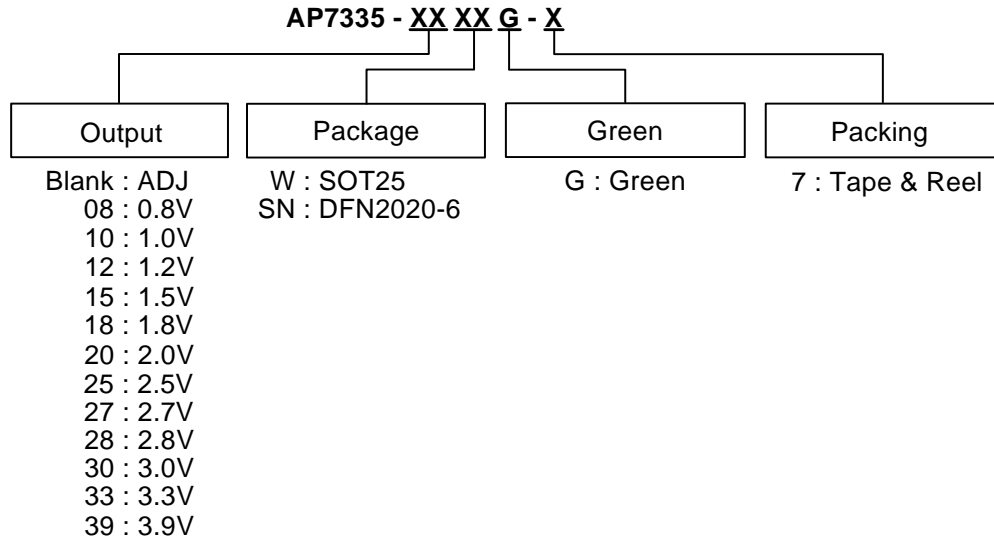
The device power dissipation and proper sizing of the thermal plane that is connected to the thermal pad is critical to avoid thermal shutdown and ensure reliable operation. Power dissipation of the device depends on input voltage and load conditions and can be calculated by:

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT}$$

The maximum power dissipation, handled by the device, depends on the maximum junction to ambient thermal resistance, maximum ambient temperature, and maximum device junction temperature, which can be calculated by the equation in the following:

$$P_D(\text{max}@T_A) = \frac{(+145^\circ\text{C} - T_A)}{R_{\theta JA}}$$

Ordering Information



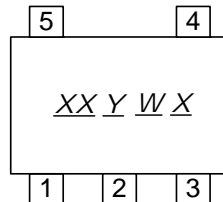
Device	Package Code	Packaging (Note 6)	7"/13" Tape and Reel	
			Quantity	Part Number Suffix
AP7335-XXWG-7	W	SOT25	3000/Tape & Reel	-7
AP7335-XXSNG-7	SN	DFN2020-6	3000/Tape & Reel	-7

Note: 6. Pad layout as shown on Diodes Inc. suggested pad layout document AP02001, which can be found on our website at <http://www.diodes.com/datasheets/ap02001.pdf>.

Marking Information

(1) SOT25

(Top View)



XX : Identification code

Y : Year 0~9

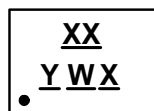
W : Week : A~Z : 1~26 week;
a~z : 27~52 week; z represents
52 and 53 week

X : A~Z : Green

Device	Package	Identification Code
AP7335-ADJ	SOT25	ZA
AP7335-08	SOT25	ZB
AP7335-10	SOT25	ZC
AP7335-12	SOT25	ZD
AP7335-15	SOT25	ZE
AP7335-18	SOT25	ZF
AP7335-20	SOT25	ZG
AP7335-25	SOT25	ZH
AP7335-27	SOT25	ZI
AP7335-28	SOT25	ZJ
AP7335-30	SOT25	ZK
AP7335-33	SOT25	ZM
AP7335-39	SOT25	ZN

(2) DFN2020-6

(Top View)



XX : Identification Code

Y : Year : 0~9

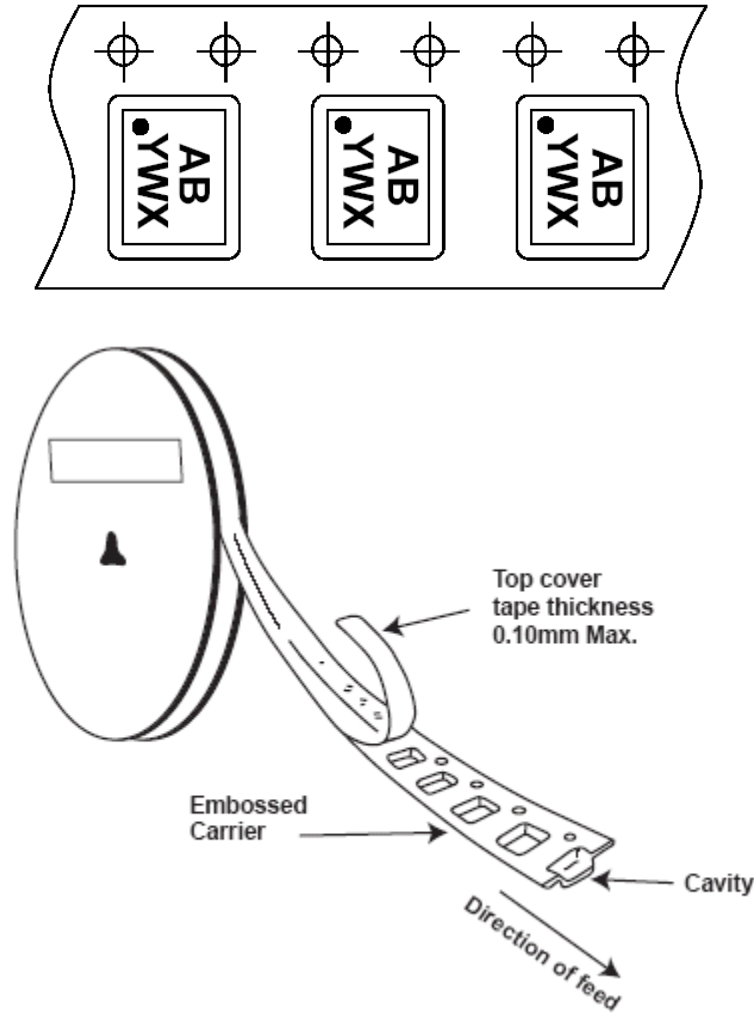
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AP7335-15	DFN2020-6	ZE
AP7335-18	DFN2020-6	ZF
AP7335-20	DFN2020-6	ZG
AP7335-25	DFN2020-6	ZH
AP7335-27	DFN2020-6	ZI
AP7335-28	DFN2020-6	ZJ
AP7335-30	DFN2020-6	ZK
AP7335-33	DFN2020-6	ZM
AP7335-39	DFN2020-6	ZN

Taping Orientation (Note 7)

For DFN2020-6



Note: 7. The taping orientation of the other package type can be found on our website at <http://www.diodes.com/datasheets/ap02007.pdf>.

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